

Optimized Design of Anti-Misalignment Magnetic Coupler for Static and Dynamic Inductive Power Transfer

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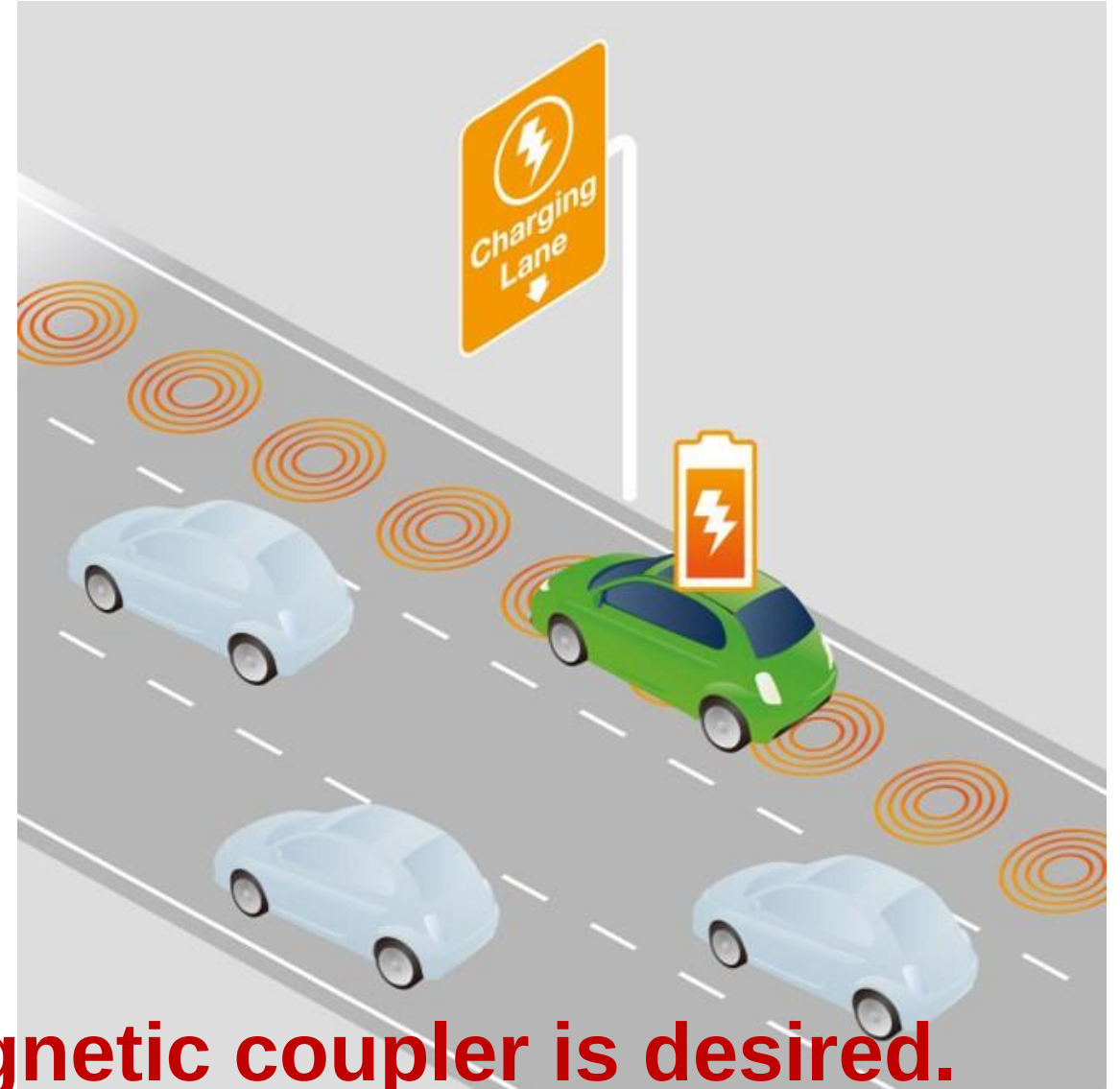
Outline

- **Research Background and Motivation**
- **Rectangular Solenoidal Pad (RSP)**
- **Control Free Design for RSP Tx**
- **Dynamic Tx Track based on RSP**

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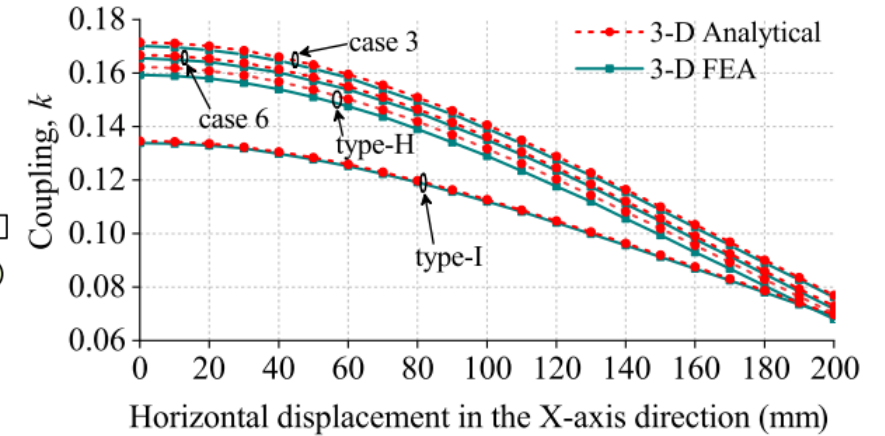
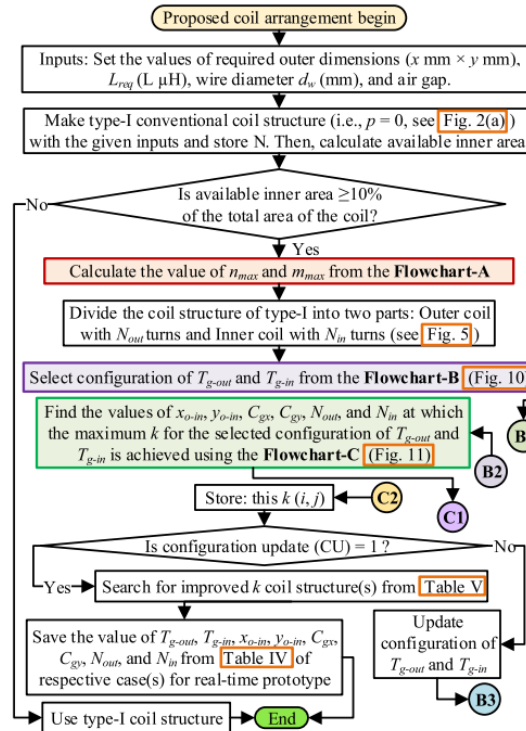
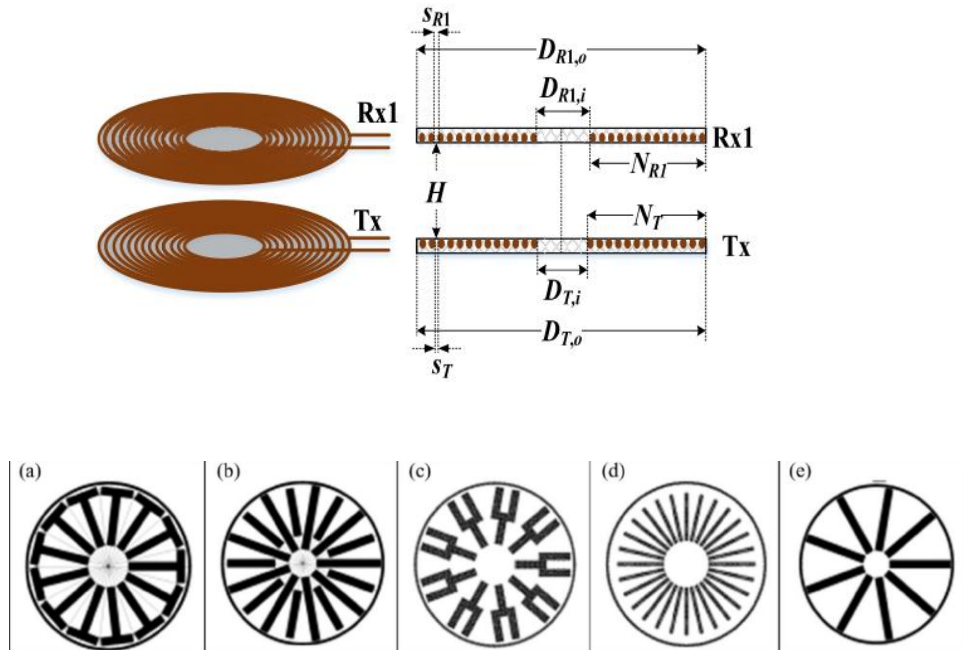
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From Static to Dynamic Wireless EV Charging



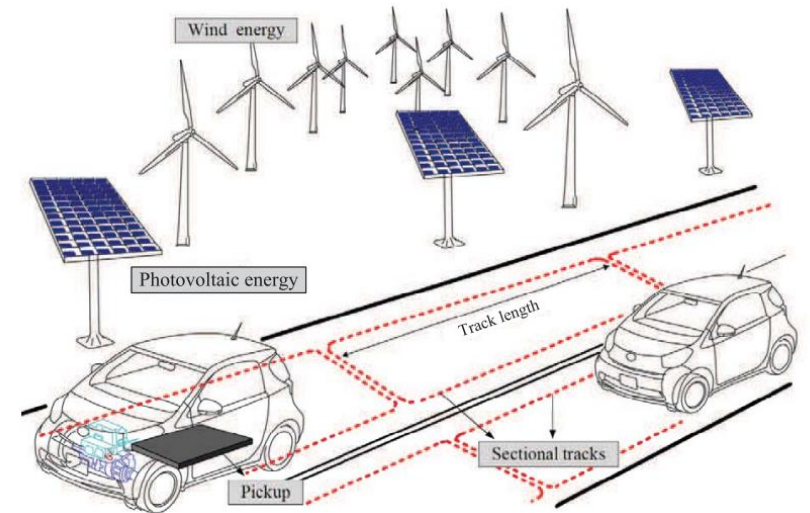
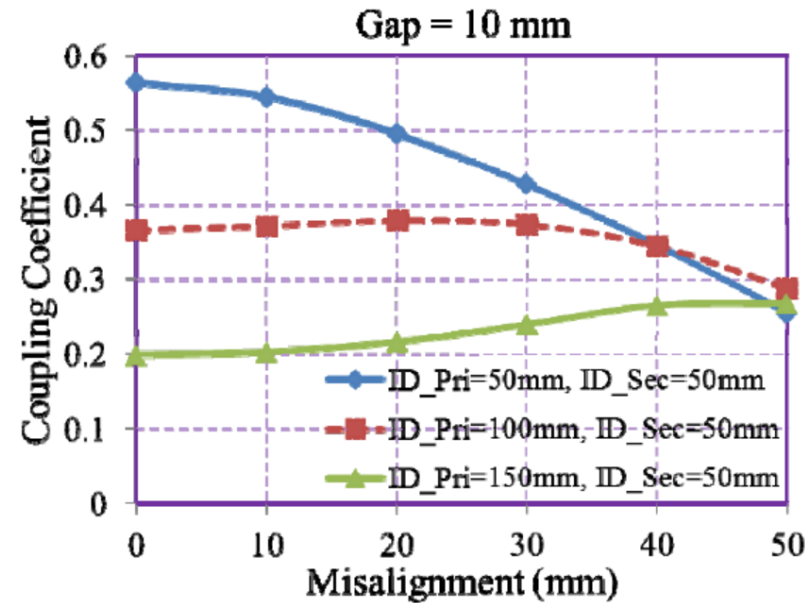
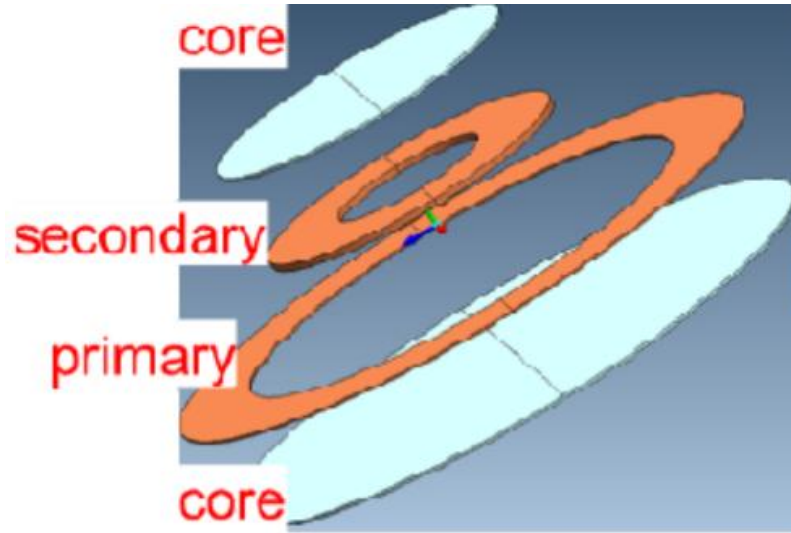
Anti-misalignment magnetic coupler is desired.

Symmetric Design



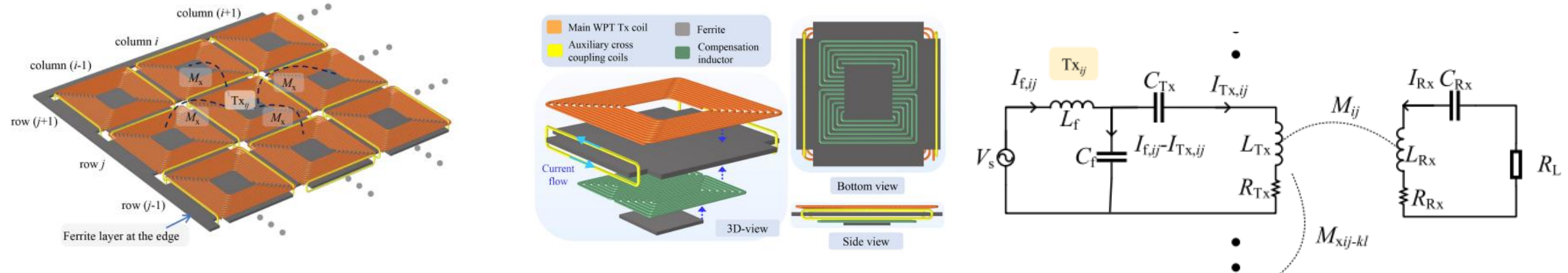
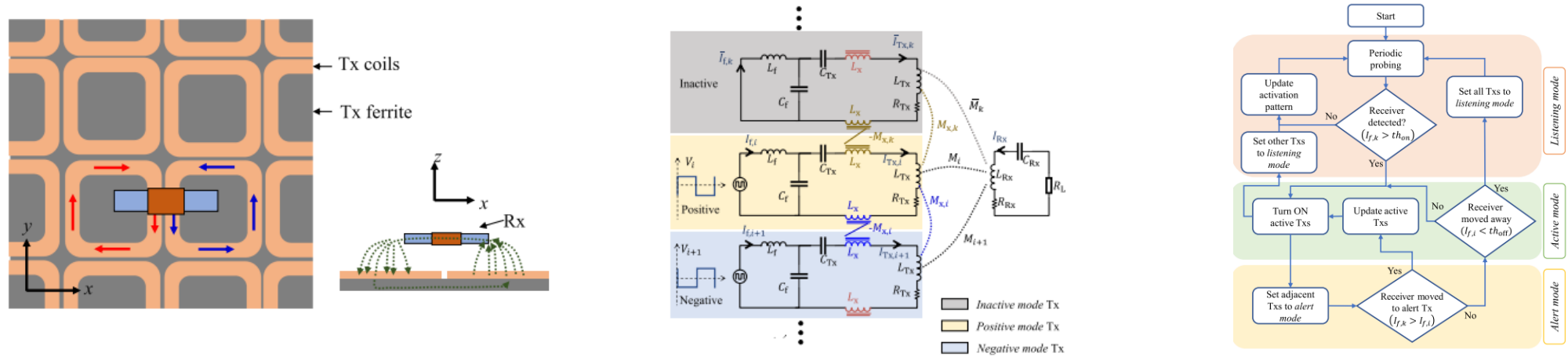
Coupling coefficient degrades significantly if misalignment occurs.

Asymmetric Design



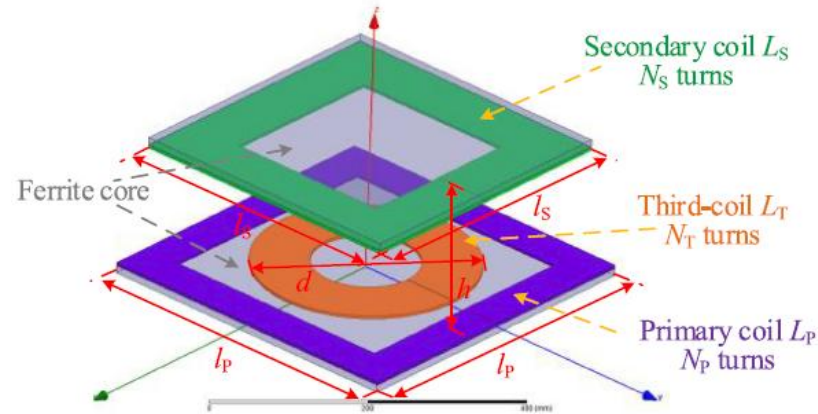
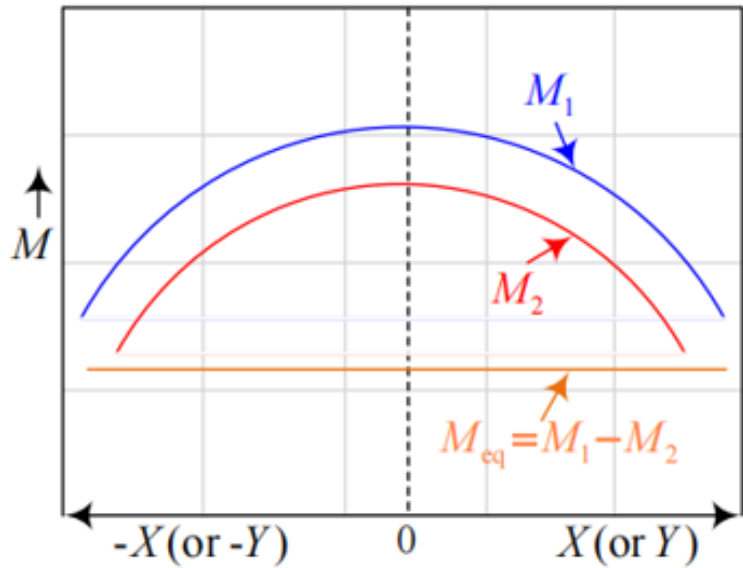
relatively even but poor magnetic coupling due to magnetic leakage

Planar Array Design

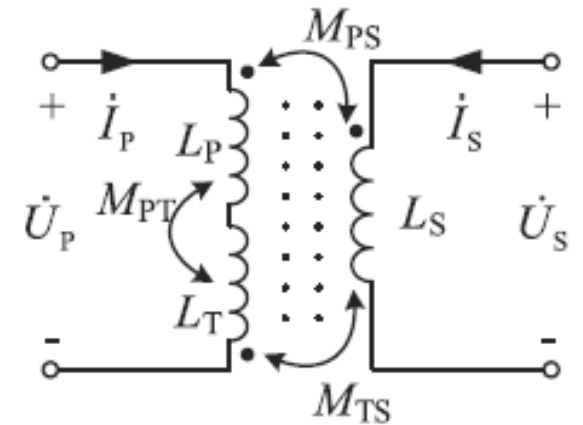


Complex control for activation and cancellation of mutual coupling

Magnetic Field Cancellation

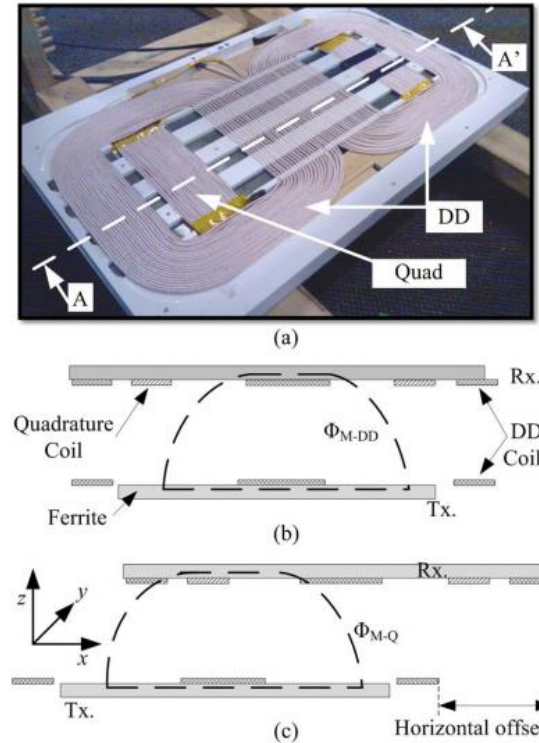
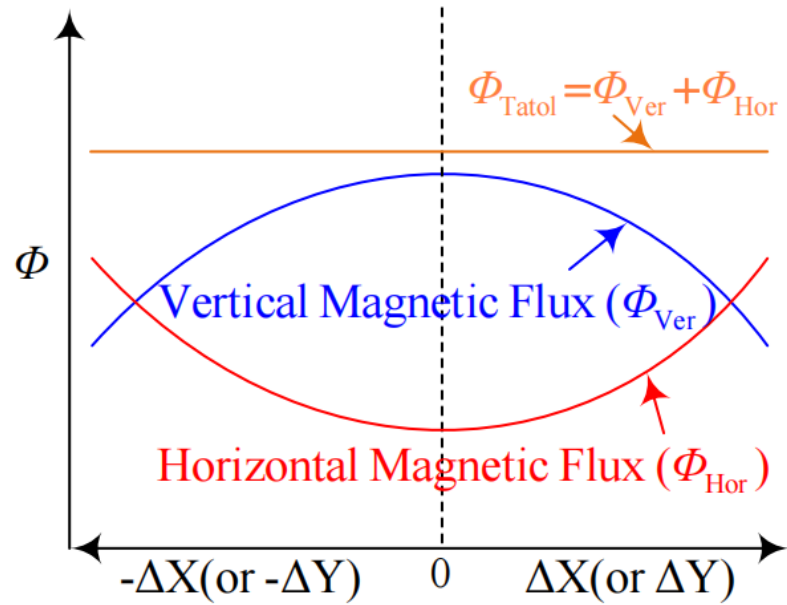


$$M_{eq} = |M_{PS}| - |M_{TS}|$$

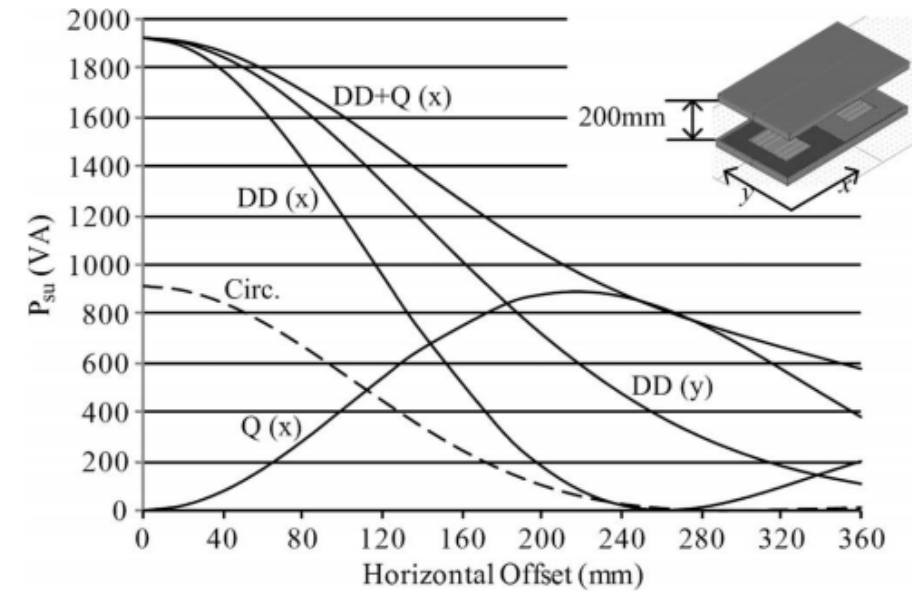


Excessive copper usage, but degraded power transfer capability

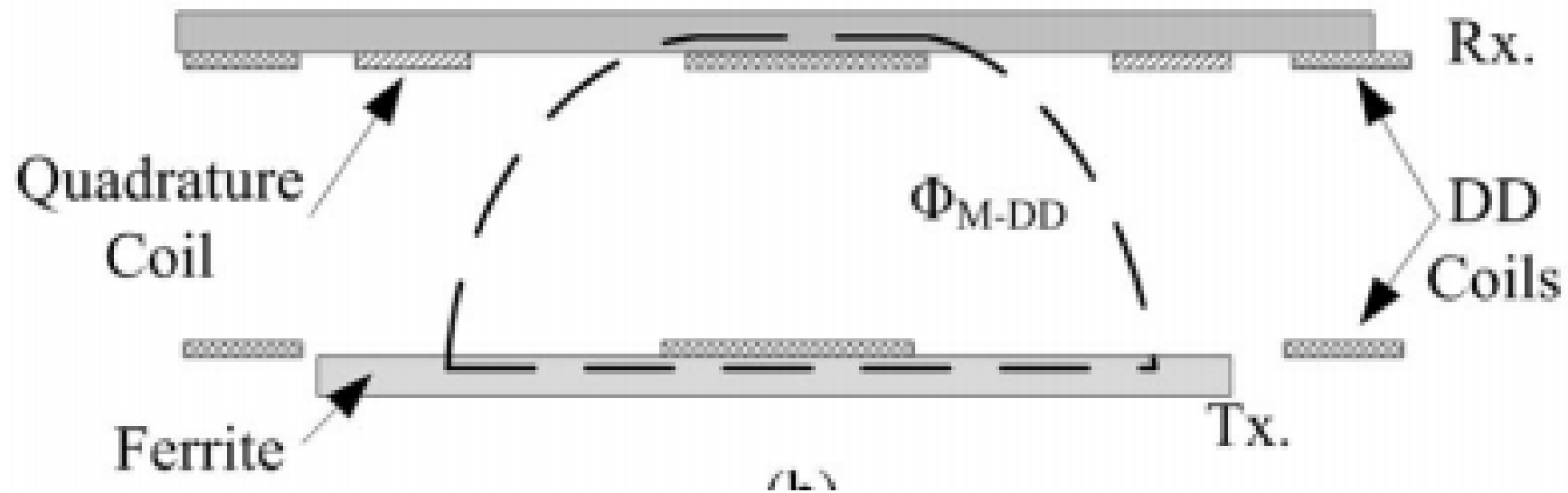
Magnetic Field Complementation



$$M_{eq} = |M_{PS}| + |M_{TS}|$$



DDQP is widely used due to its good anti-misalignment capability.

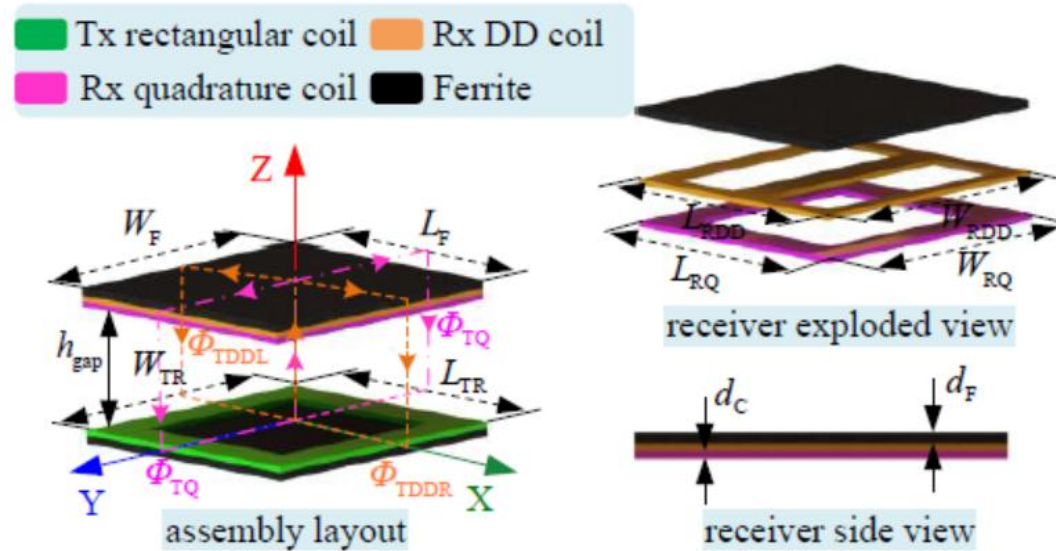


Is the DDQP Good Enough?

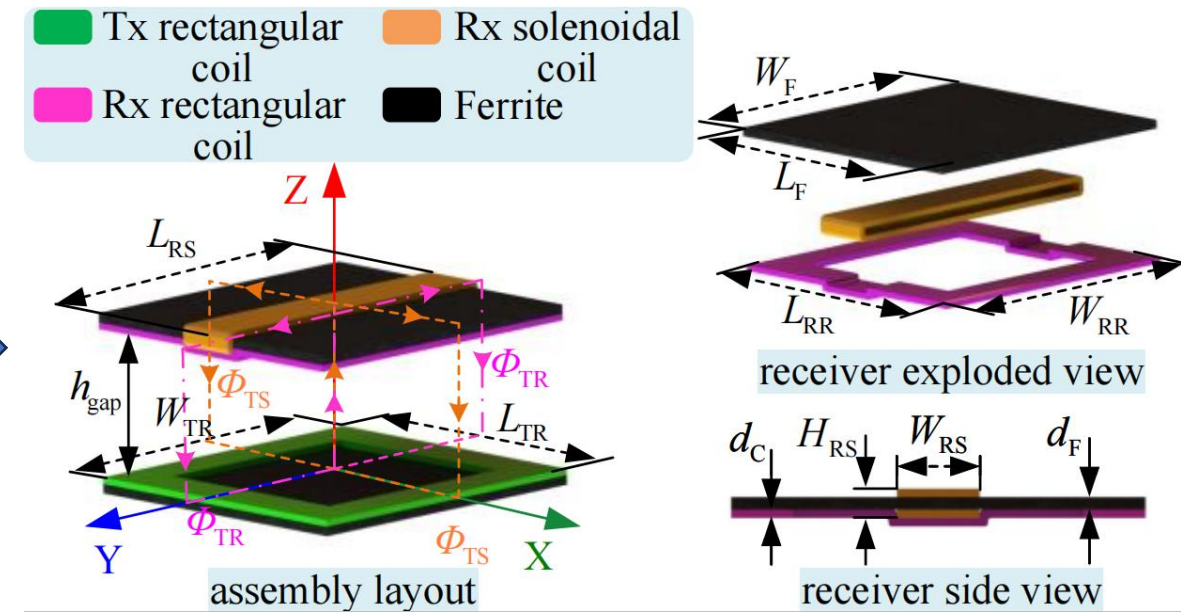
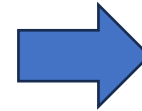
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From DD coil to Solenoidal Coil

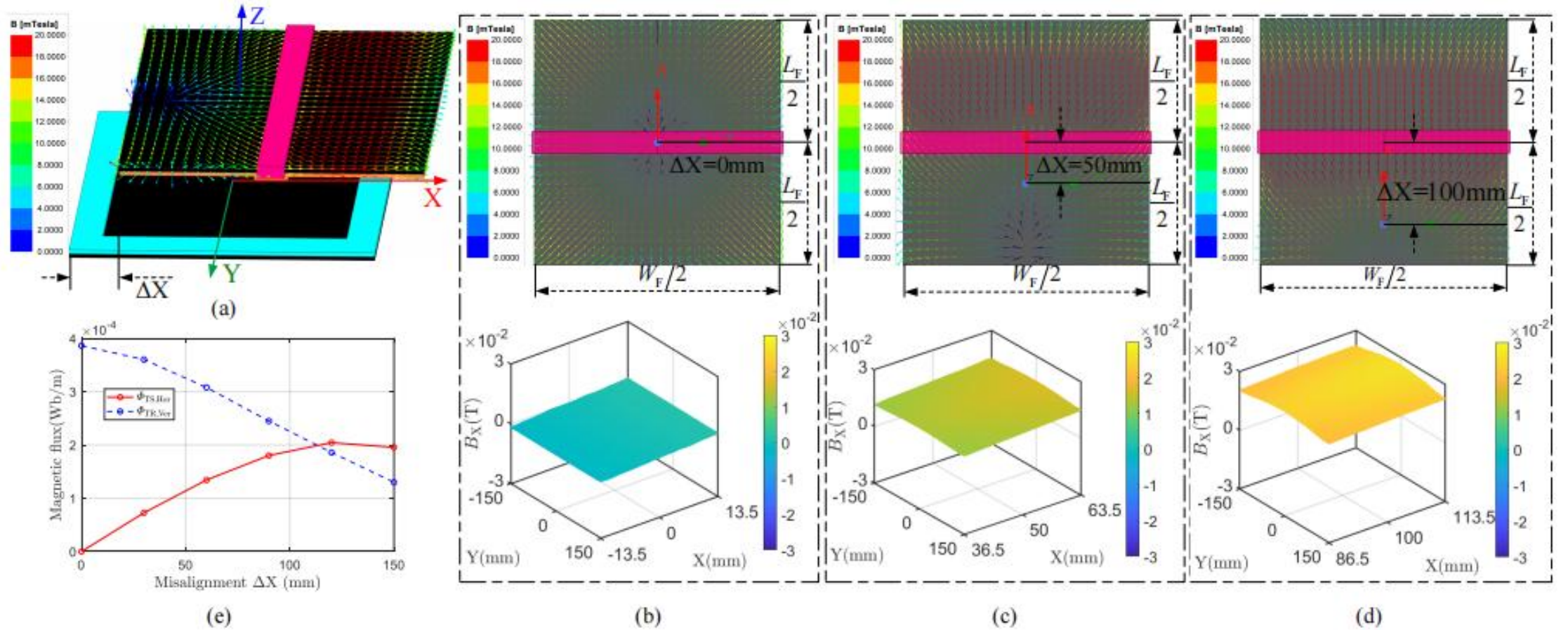


RP-DDQP



RP-RSP

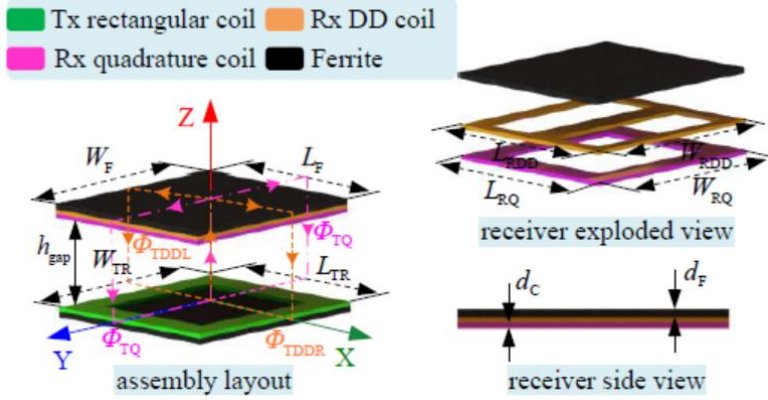
Operating Principle



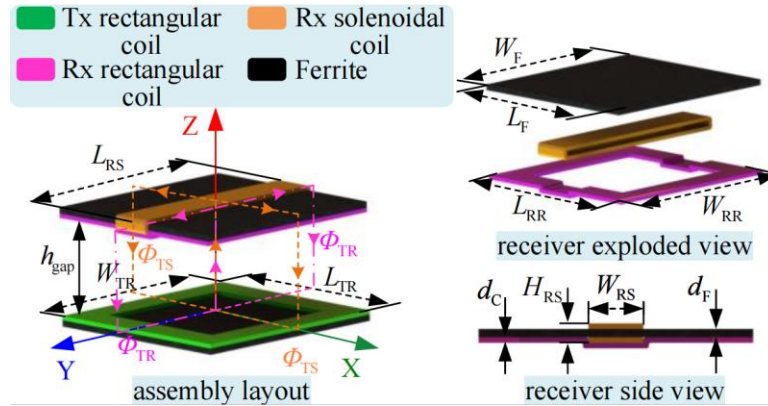
(a) Coordinates. (b) $\Delta X = 0$ mm (fully aligned). (c) $\Delta X = 50$ mm. (d) $\Delta X = 100$ mm. (e) Diagrams of $\phi_{TR,Ver}$ and $\phi_{TS,Hor}$

Capture vertical and horizontal magnetic field in a complementary manner for pad misalignment

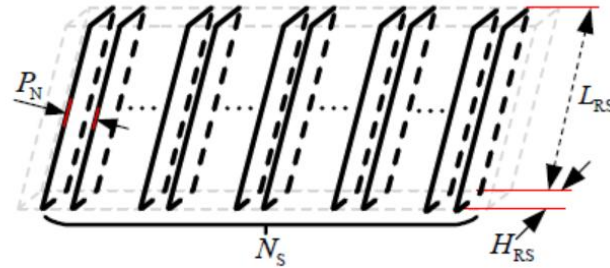
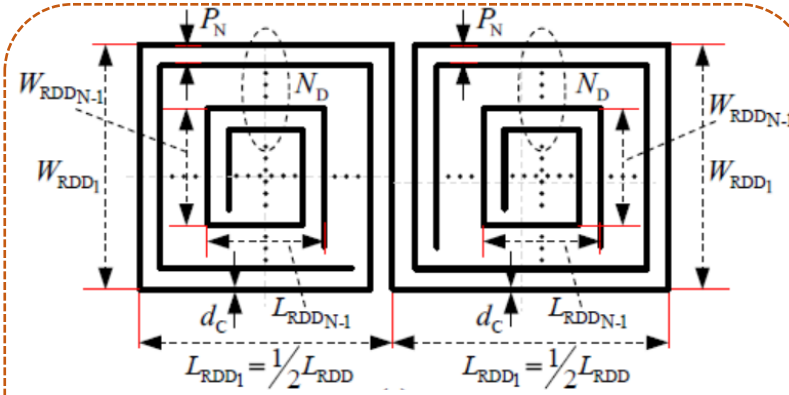
Magnetic Flux Capture Capability



DDQP



Rectangular-Solenoidal Pad (RSP)



Effective magnetic flux captured by DD coil with N_D turns

$$\Phi_{TDD} = \int_{S_{L_1}} B_{Z_{L_1}} \cos \alpha dS + \cdots + \int_{S_{L_{N_D}}} B_{Z_{L_{N_D}}} \cos \alpha dS \\ (- \int_{S_{R_1}} B_{Z_{R_1}} \cos \alpha dS + \cdots + \int_{S_{R_{N_D}}} B_{Z_{R_{N_D}}} \cos \alpha dS).$$

Make $S_{DD_1} = S_{L_1} = S_{R_1}$.

where

$$S_{DD_1} = \frac{1}{2} L_{RDD} W_{RDD}$$

$$S_{DD_2} = (\frac{1}{2} L_{RDD} - 2(d_C + P_N))(W_{RDD} - 2(d_C + P_N))$$

...

$$S_{DD_N} = (\frac{1}{2} L_{RDD} - 2(N_D - 1)(d_C + P_N))(W_{RDD} - 2(N_D - 1)(d_C + P_N))$$

Effective magnetic flux captured by solenoidal coil with N_S turns

$$\phi_{TS} = \int_{S_1} B_{X_{S_1}} \cos \alpha dS + \int_{S_2} B_{X_{S_2}} \cos \alpha dS + \cdots + \int_{S_i} B_{X_{S_i}} \cos \alpha dS + \cdots + \int_{S_{N_S}} B_{X_{S_{N_S}}} \cos \alpha dS.$$

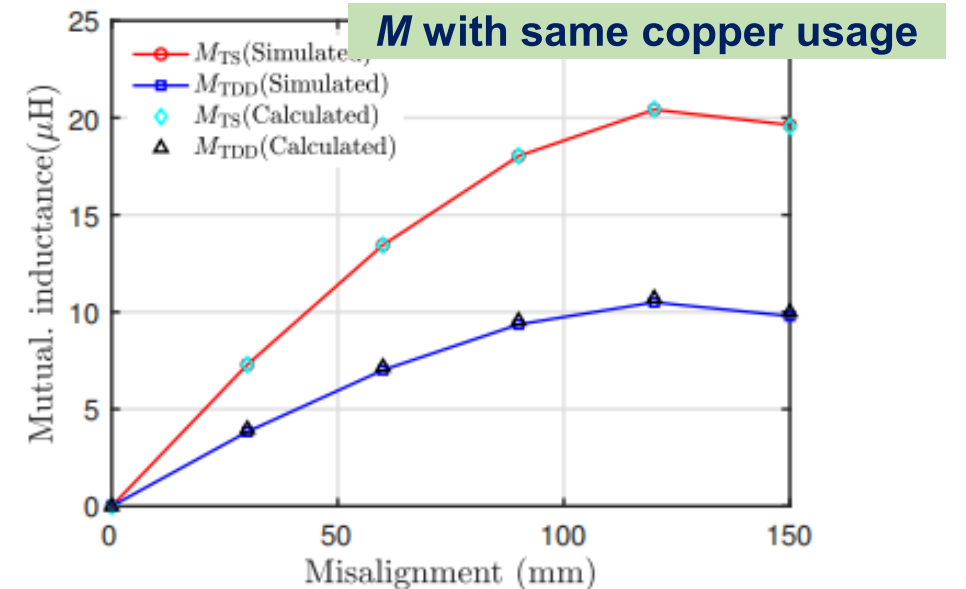
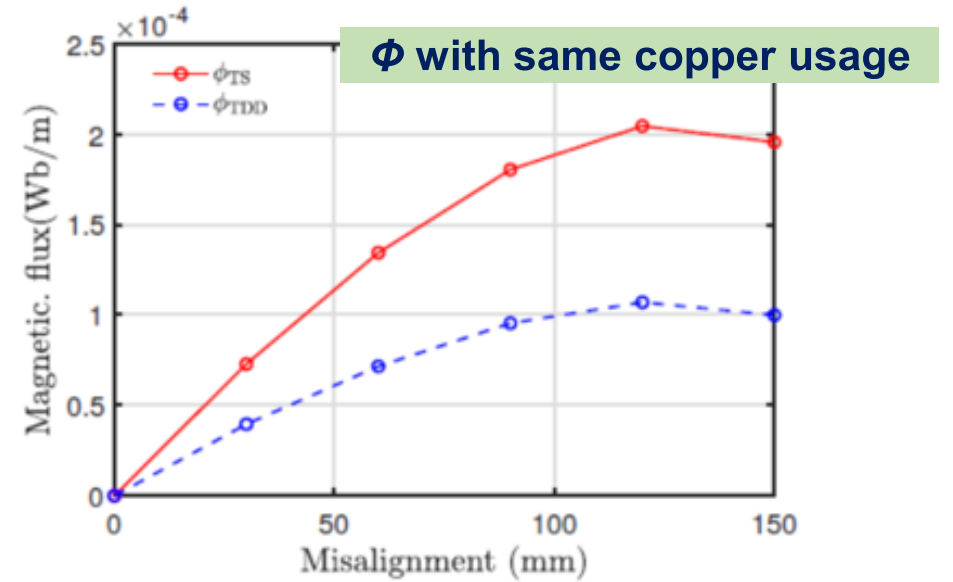
$$S_S = S_1 = S_2 = \cdots = S_{N_S} = (L_{RS} - 2d_C)(H_{RS} - 2d_C).$$

Magnetic Flux Capture Capability

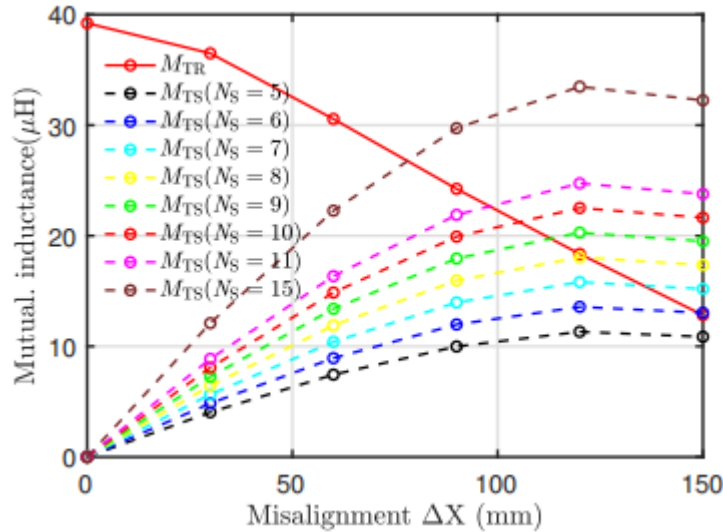
TABLE I
PARAMETERS OF THE COUPLER

Items	Symbols	Parameters	Value(mm)
RSP coupler	L_{RR}	Lenght of Rx rectangular coil	300
	W_{RR}	Width of Rx rectangular coil	300
	L_{RS}	Lenght of Rx solenoidal coil	306
	H_{RS}	Height of Rx solenoidal coil	9
DDQP coupler	L_{RDD}	Lenght of Rx DD coil	300
	W_{RDD}	Width of Rx DD coil	300
	L_{RQ}	Lenght of Rx solenoidal coil	300
	W_{RQ}	Width of Rx quadrature coil	300
Common parameters	L_{TR}	Lenght of Tx rectangular coil	300
	W_{TR}	Width of Tx rectangular coil	300
	L_F, W_F	Ferrite dimensions	300,300
	d_F		3
	d_C	Litz wire diameter of coils	3
	h_{gap}	Power transfer gap distance	60
	N	Turn numbers of unipolar coils	10
	P_N	Turn space of coils	0

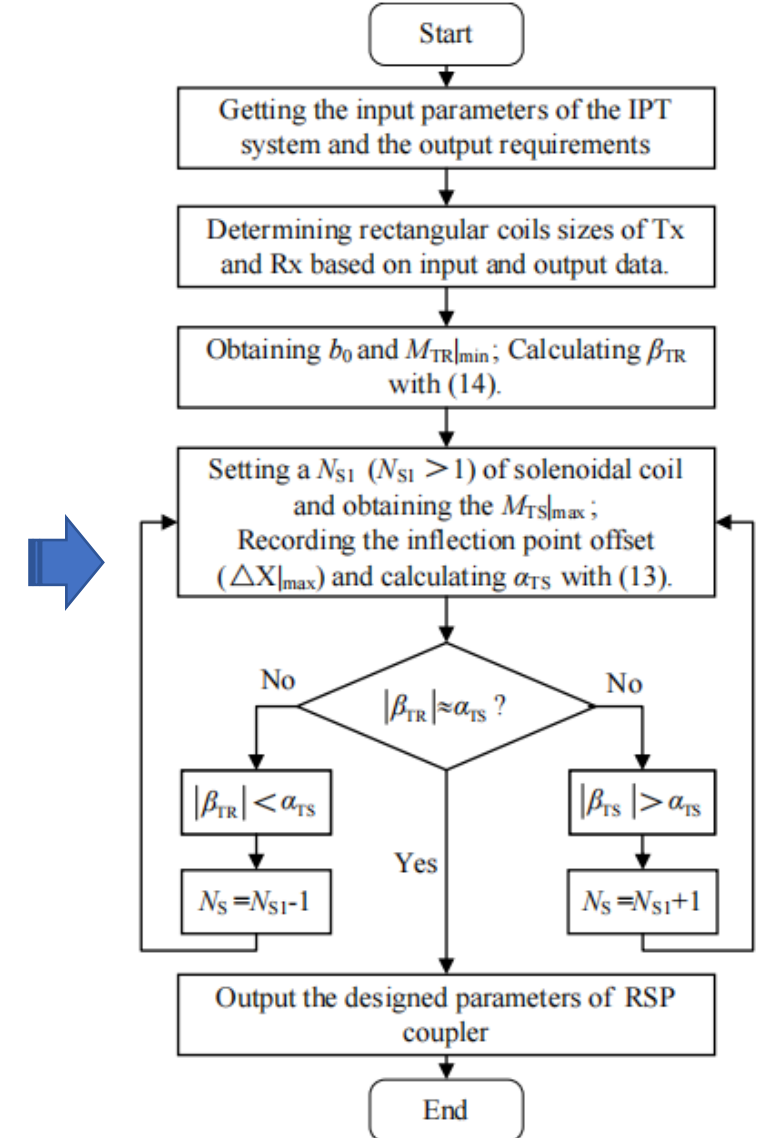
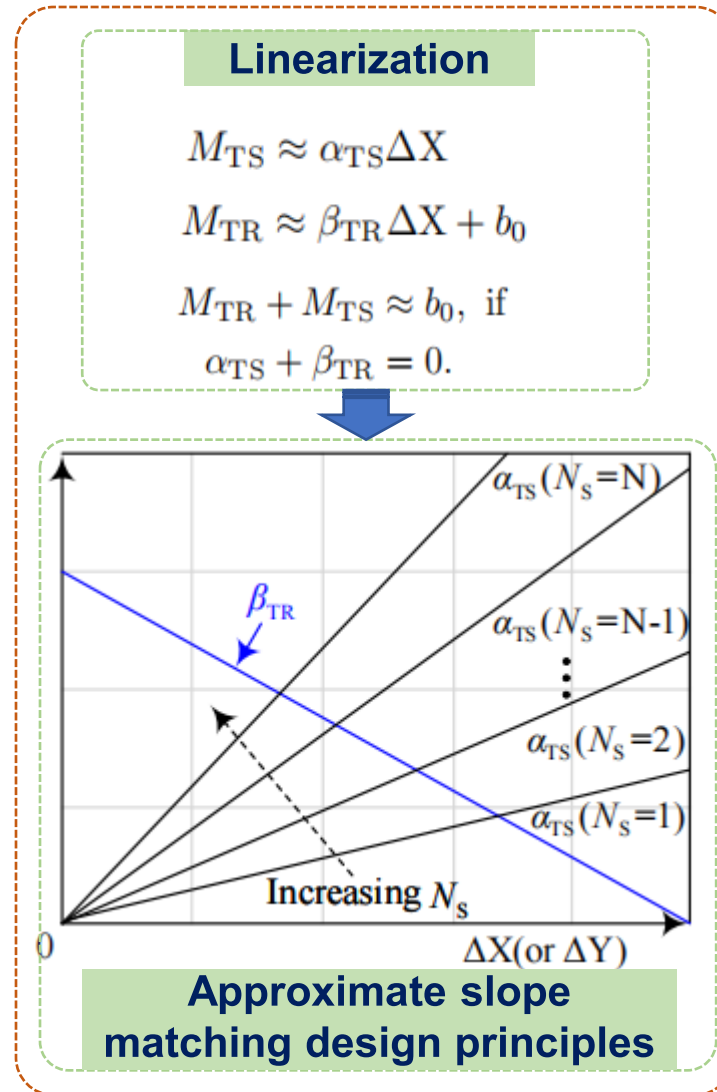
The solenoidal coil can better capture the horizontal magnetic field using the same amount of copper.



Design of Rectangular Solenoidal Coil



Mutual inductances M_{TR} and M_{TS}

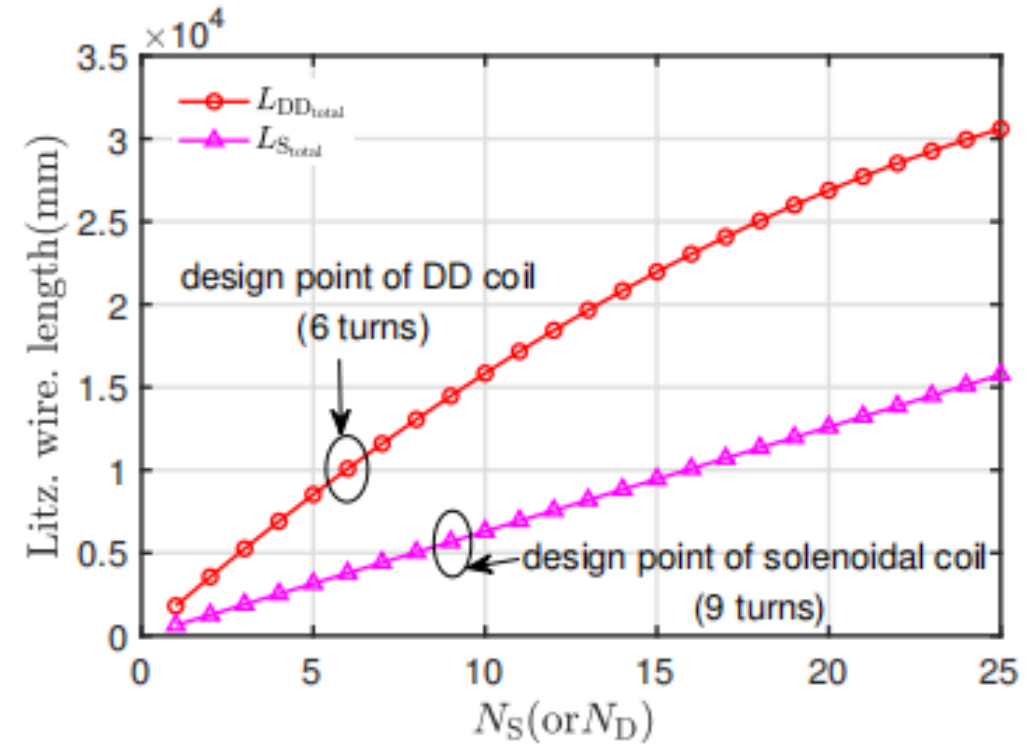


Misalignment enhancement design flowchart based on approximate slope prediction of RSP coupler

Comparison of Copper Usage

TABLE II
SLOPE AND COPPER CONSUMPTION OF RSP COUPLER AND DDQP
COUPLER VERSUS NUMBER OF TURNS

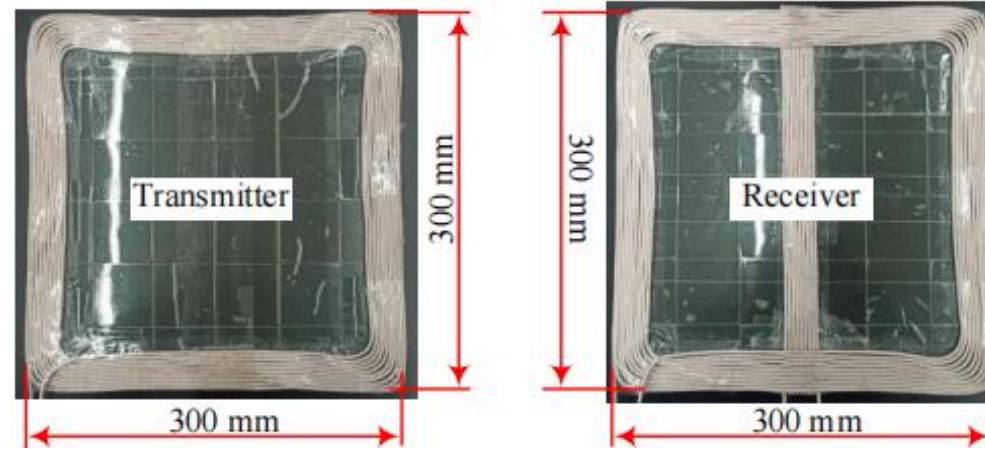
Symbol	Turns number	Slopes	Copper (mm)
Rectangular coil	10	-0.174	
	5	0.094	3150
	6	0.113	3780
	7	0.132	4410
Solenoidal coil	8	0.150	5040
	9	0.169	5670
	10	0.188	6300
	11	0.206	6930
	15	0.279	9450
	5	0.139	8520
DD coil	6	0.163	10080
	7	0.186	11592
	8	0.208	13056
	9	0.228	14472
	10	0.248	15840
	11	0.265	17160
	15	0.320	21960



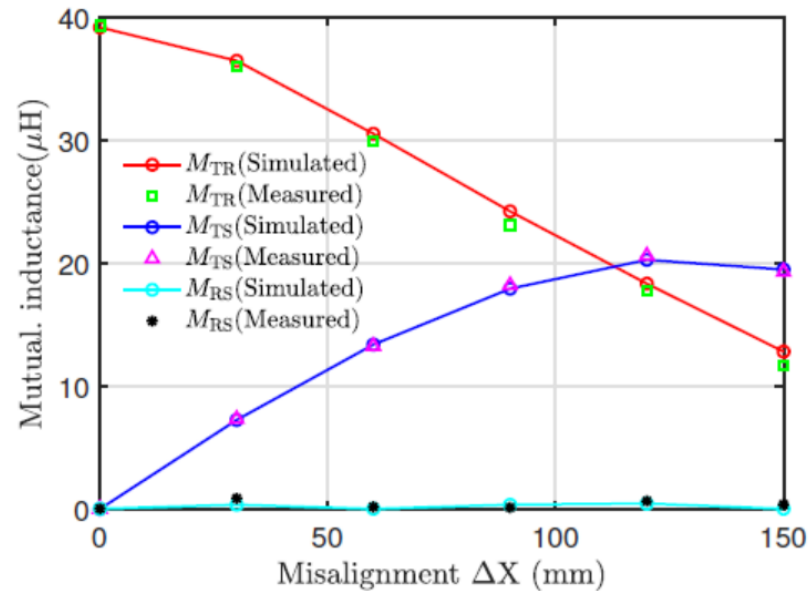
The proposed solenoidal coil can save **43.73%** litz wire consumption compared with the DD coil for the same anti-misalignment capability.

Experimental Validation

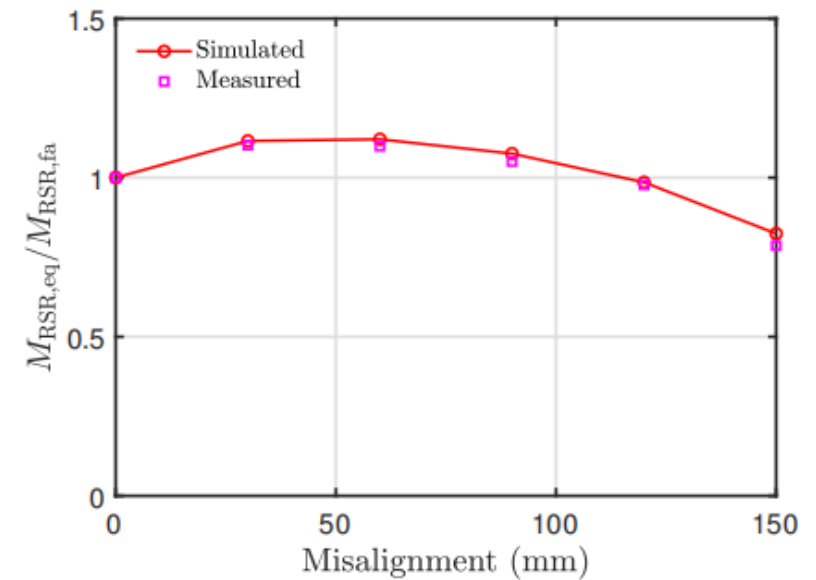
Rectangular Pad –
Rectangular Solenoidal Pad



Mutual inductances

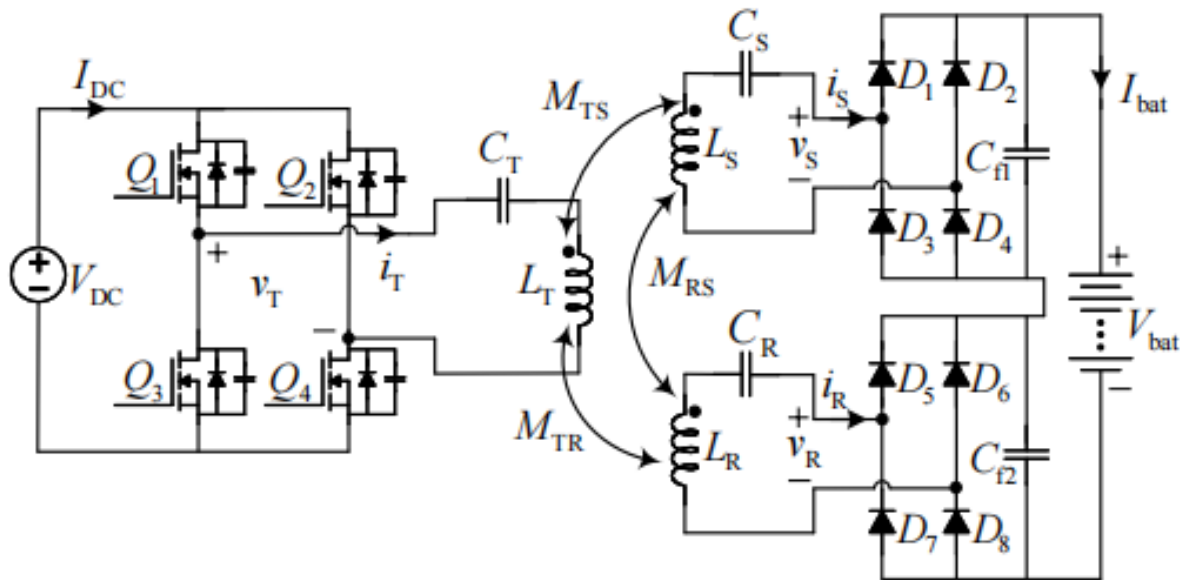


Equivalent mutual inductance

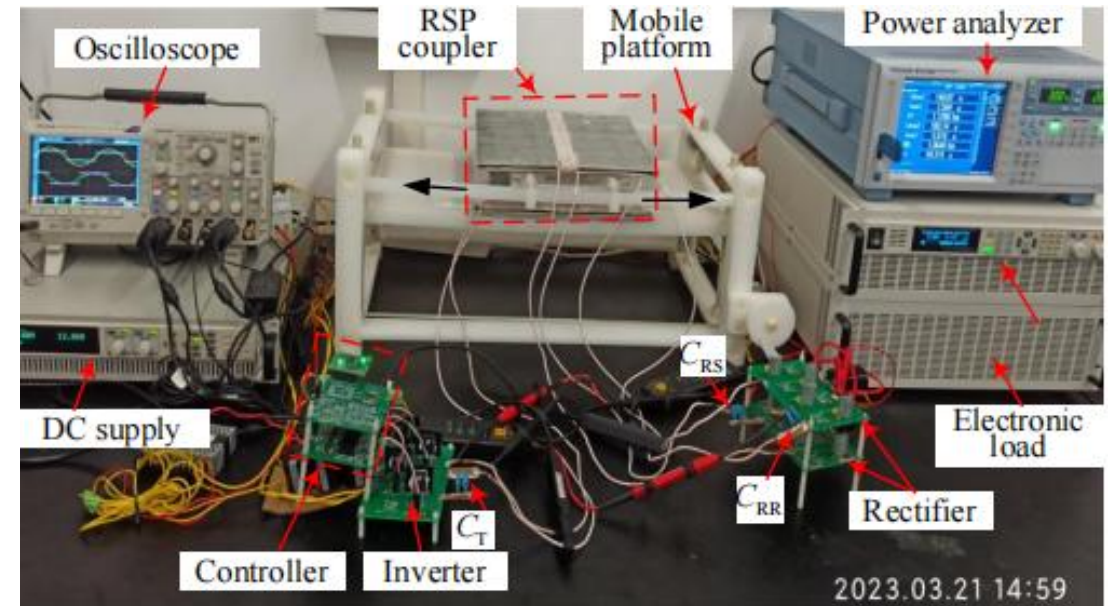


Experimental Validation

IPT system with RSP

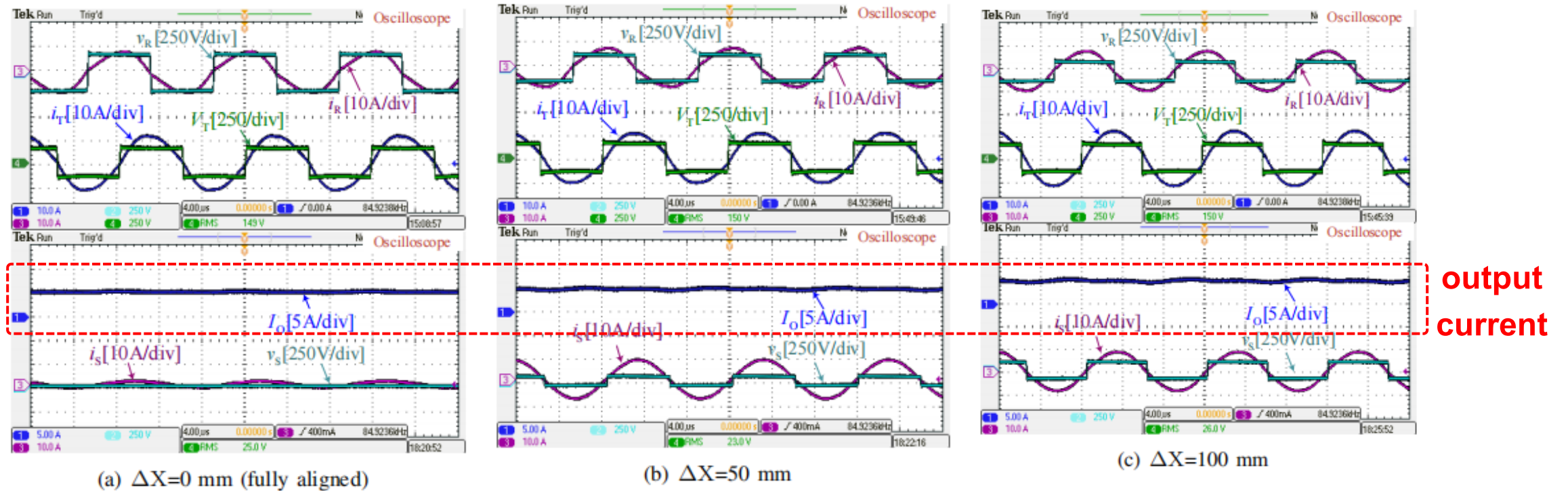


Experimental prototype of 1-kW IPT system with RSP



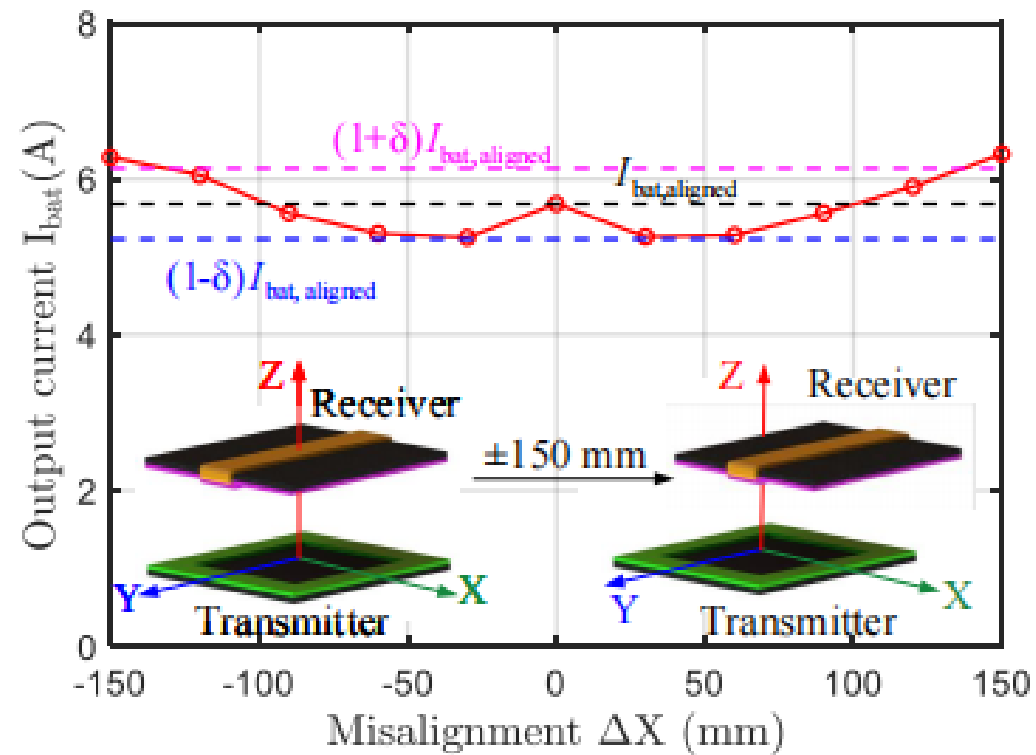
Experimental Validation

Experimental waveforms of the proposed IPT charging system with RSP at various misalignment conditions

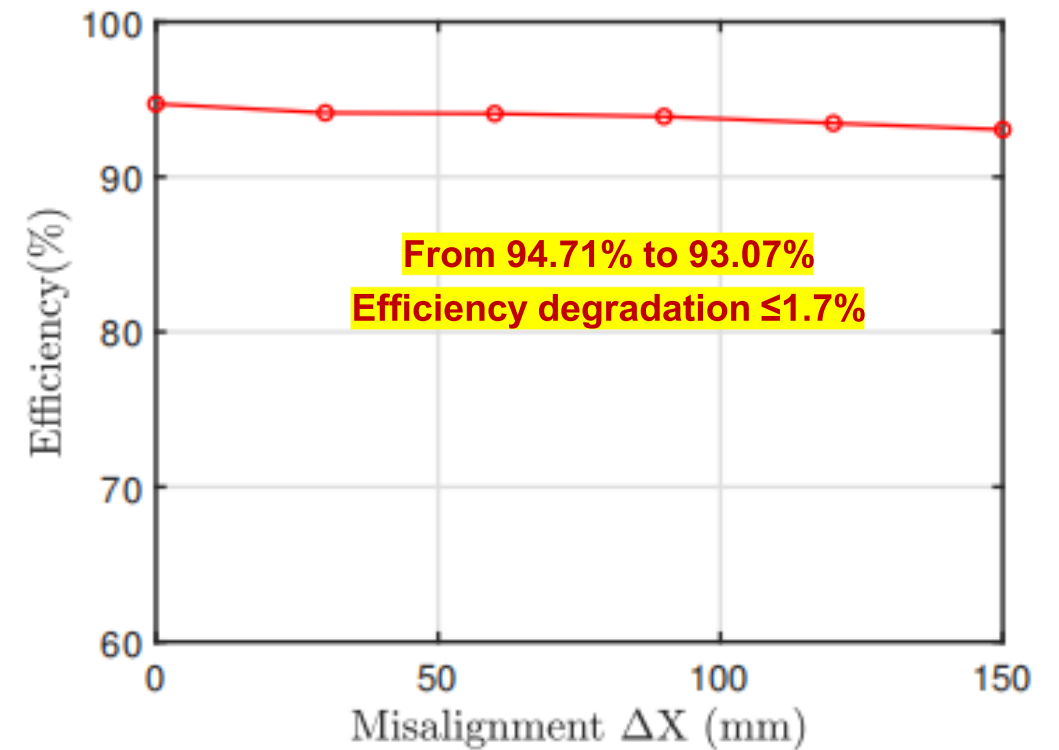


Experimental Validation

Output current under different misalignments along X-axis



Measured transfer efficiency under different misalignments along the X-axis

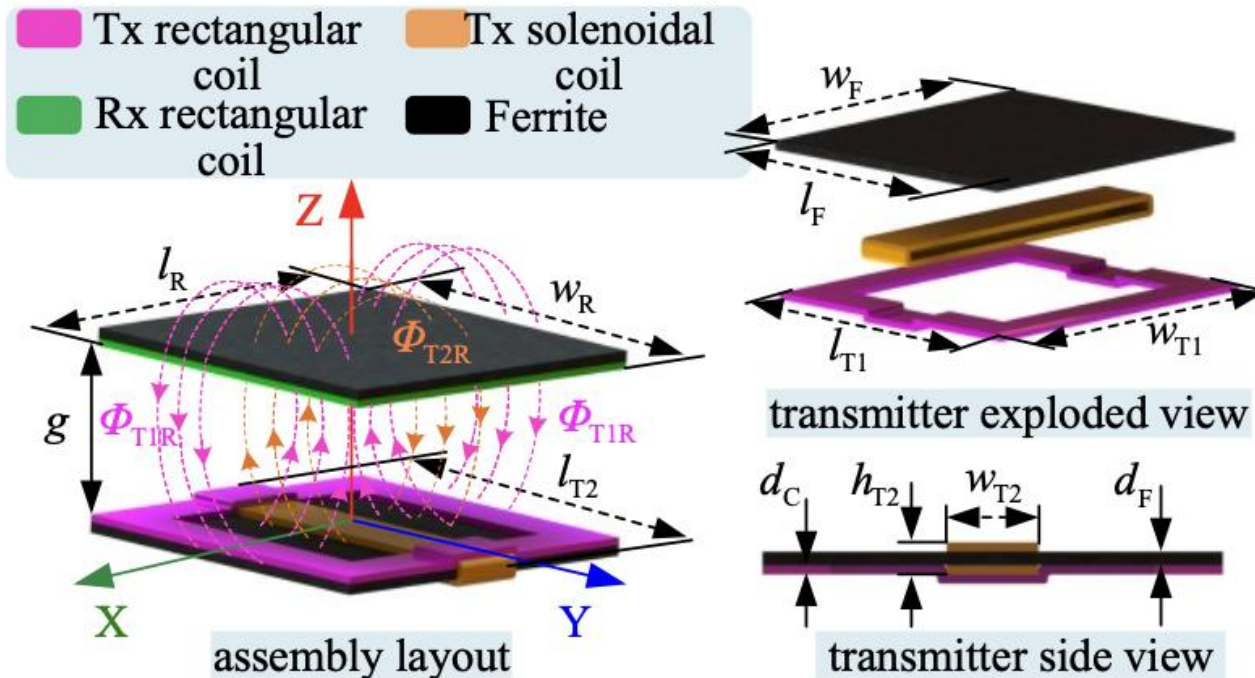


Outline

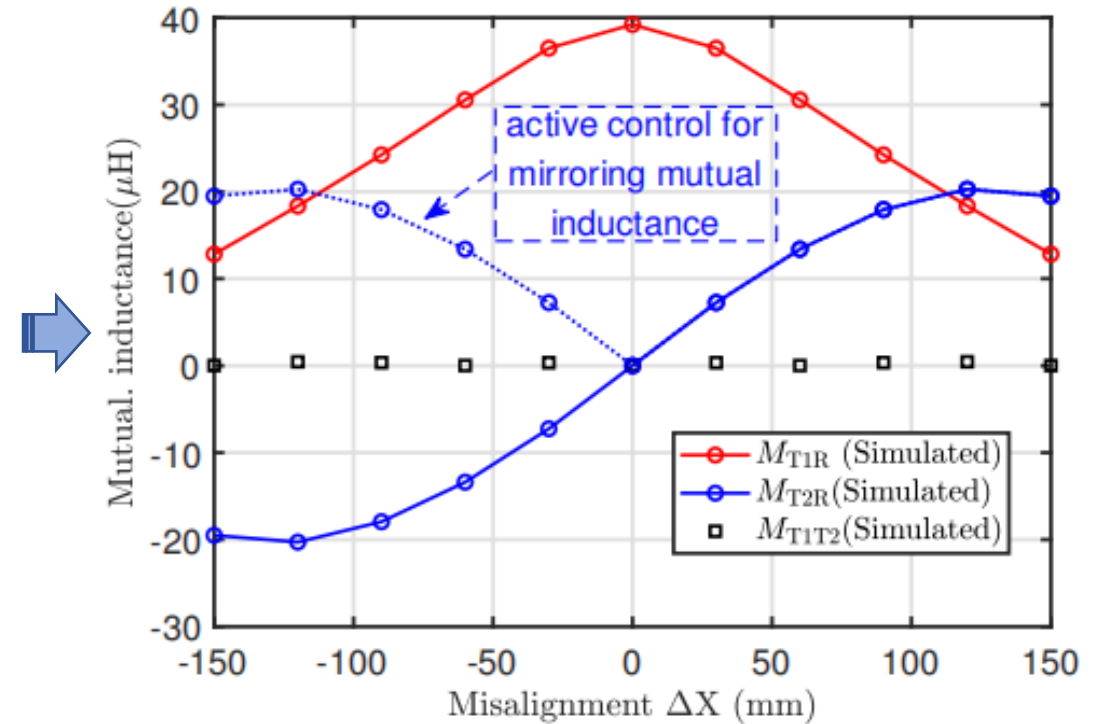
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Polarity Reversal Problem

RSP-RP coupler

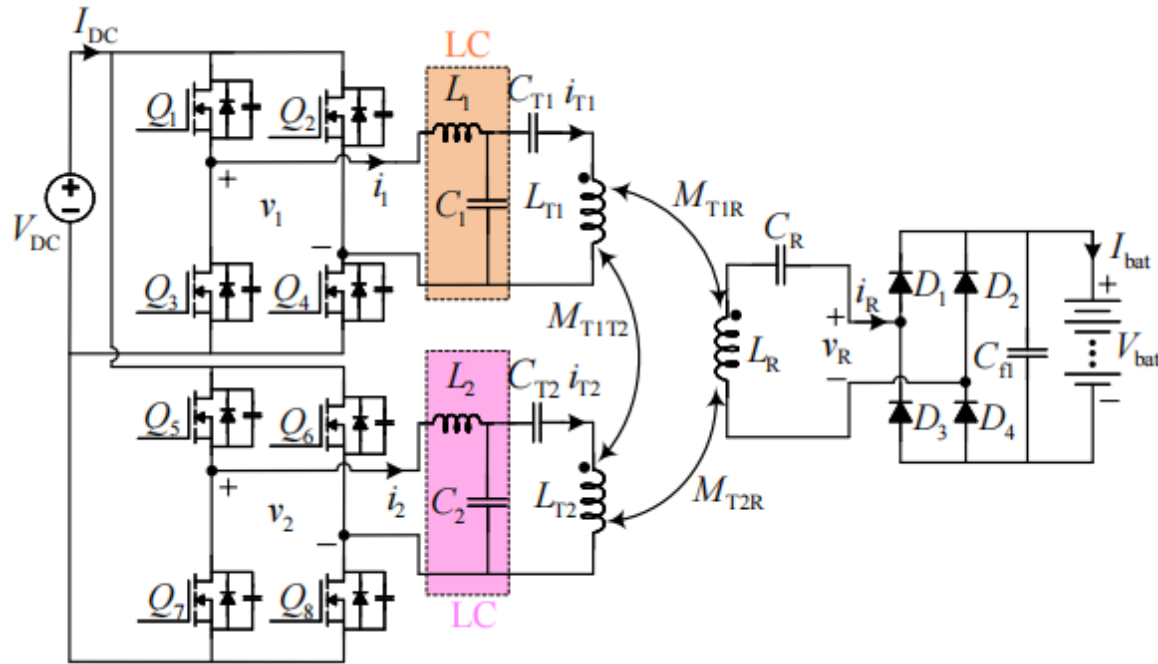


Mutual inductances of RSP-RP versus misalignments along X-axis



Can detection and control be eliminated?

Circuit Modeling



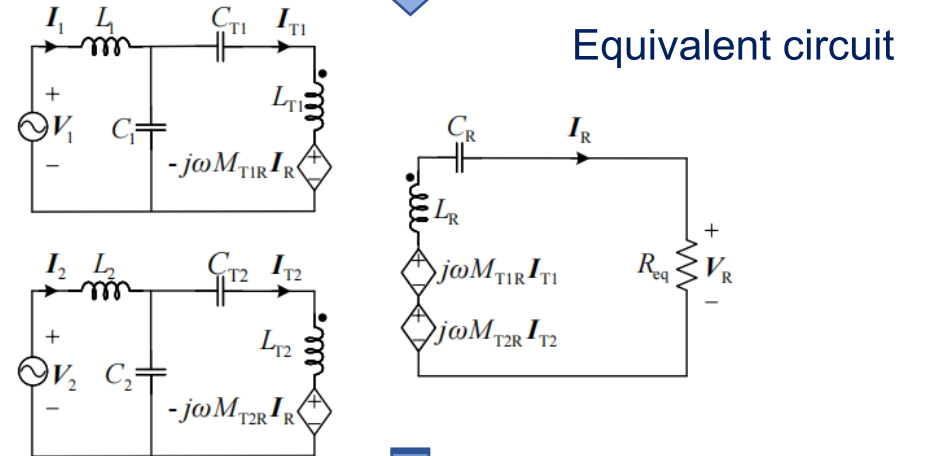
Proposed IPT charging system with RSP-RP coupler



$$\omega L_1 - \frac{1}{\omega C_1} = \omega L_2 - \frac{1}{\omega C_2} = \omega L_R - \frac{1}{\omega C_R} = 0. \quad (4)$$

$$I_{T1} = \frac{V_1}{j \frac{1}{\omega C_1}}, \text{ and} \quad (5)$$

$$I_{T2} = \frac{V_2}{j \frac{1}{\omega C_2}} = \frac{V_1 \angle \theta}{j \frac{1}{\omega C_2}}, \quad (6)$$



$$0 = \left(\frac{1}{j\omega C_{T1}} + j\omega L_{T1} \right) I_{T1} - j\omega M_{T1R} I_R, \quad (7)$$

$$0 = \left(\frac{1}{j\omega C_{T2}} + j\omega L_{T2} \right) I_{T2} - j\omega M_{T2R} I_R,$$

$$V_R = R_{eq} I_R = j\omega M_{T1R} I_{T1} + j\omega M_{T2R} I_{T2}.$$

With (7), output voltage V_R is given by

$$V_R = \omega^2 C_1 M_{T1R} V_1 + \omega^2 C_2 M_{T2R} V_1 \angle \theta. \quad (8)$$

Analysis of Output Function

The amplitude of V_R is given by (9).

$$|V_R| = \omega^2 C_1 |V_1| \beta, \quad (9)$$

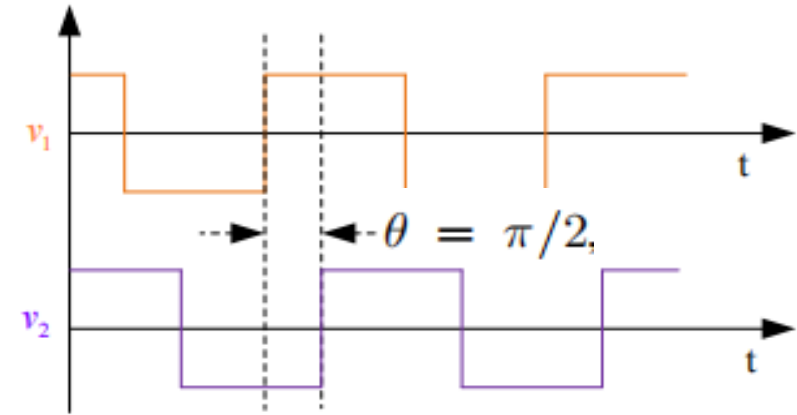
where

$$\beta = \sqrt{\underbrace{\frac{2C_2 M_{T1R} M_{T2R} \cos\theta}{C_1}}_{\text{Term ①}} + \underbrace{M_{T1R}^2 + \frac{C_2^2}{C_1^2} M_{T2R}^2}_{\text{Term ②}}} \quad (10)$$

Term ① = 0 ↑

↑ **Term ② ≈ constant**

$$\lambda = \frac{\Delta|_{\text{Max}} - \Delta|_{\text{Min}}}{\Delta|_{\text{Max}} + \Delta|_{\text{Min}}} \times 100\%, \quad (13)$$



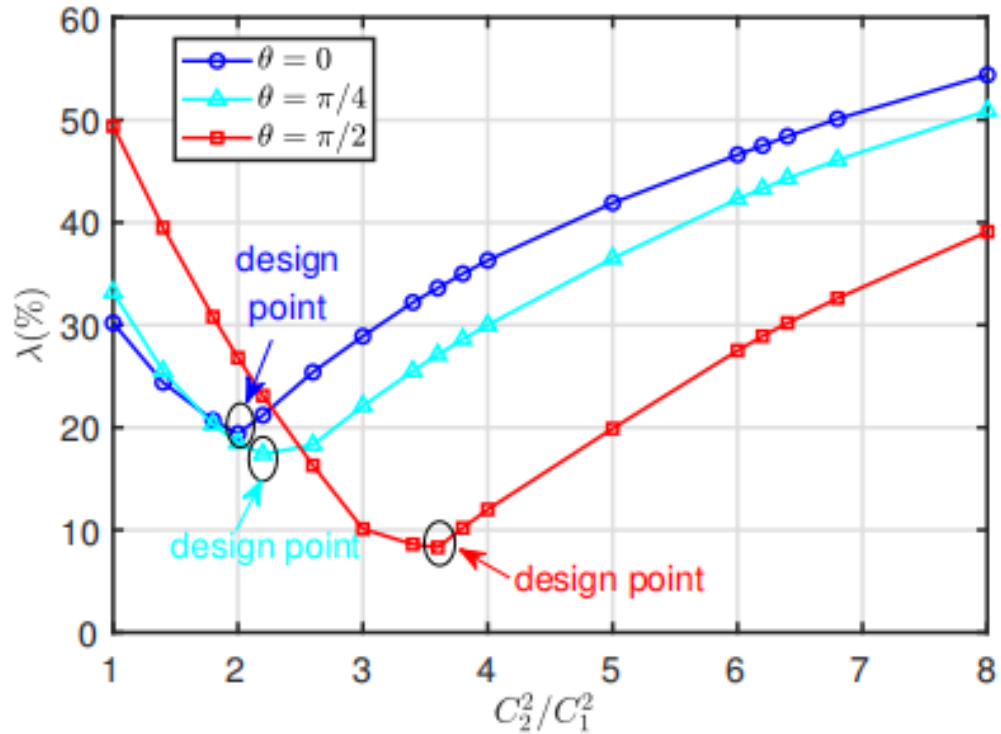
$$V_O = \omega^2 V_{DC} C_1 \sqrt{M_{T1R}^2 + \frac{C_2^2}{C_1^2} M_{T2R}^2}. \quad (11)$$

$$\Delta = M_{T1R}^2 + \frac{C_2^2}{C_1^2} M_{T2R}^2. \quad (12)$$

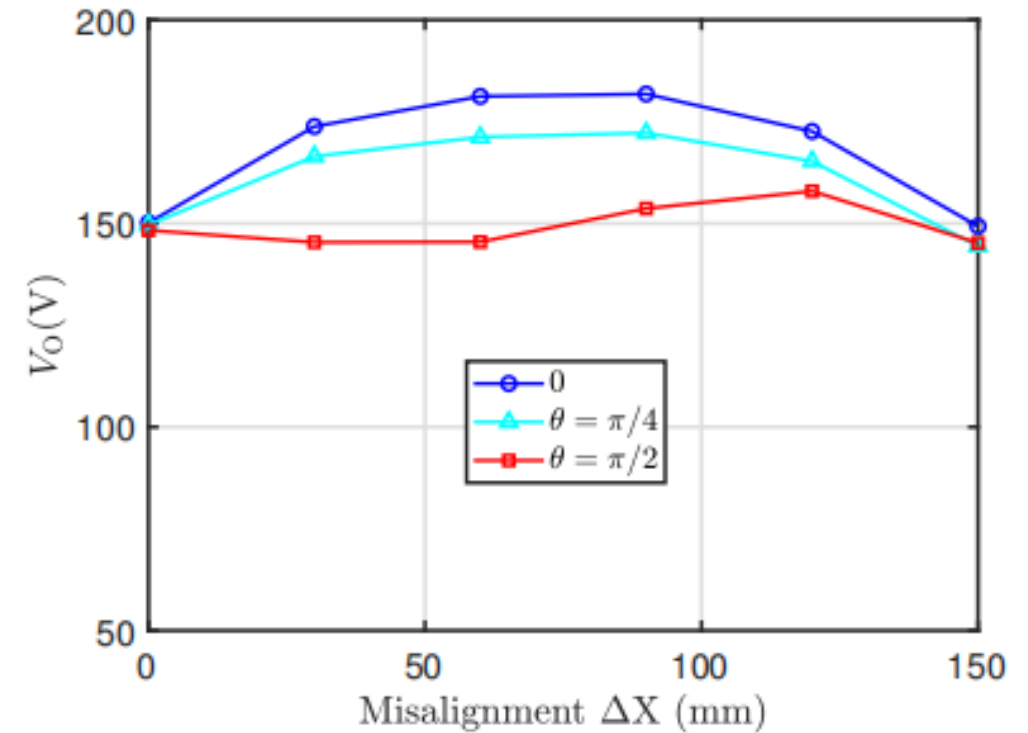


Orthogonal excitation

Design for Minimal Output Fluctuation

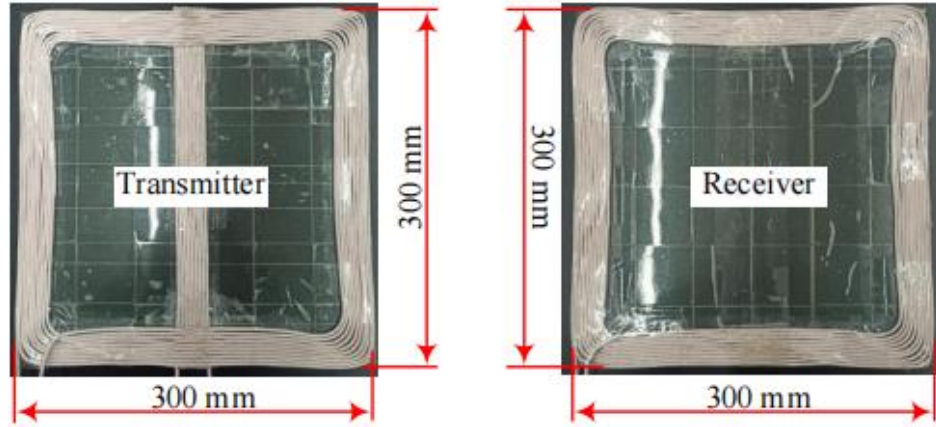


Fluctuation λ versus C_2^2/C_1^2 under different phase difference θ of excitation currents.

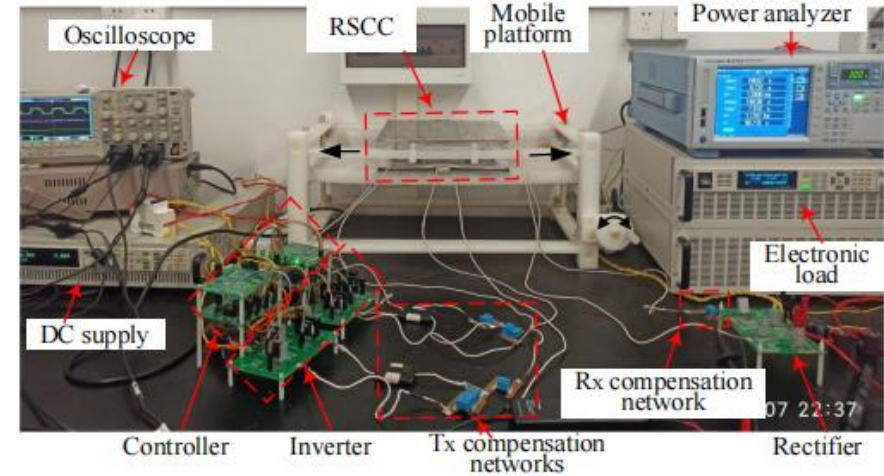


Output voltage versus misalignment along X-axis under different phase difference θ of excitation currents.

Experimental Validation



RSP-RP coupler



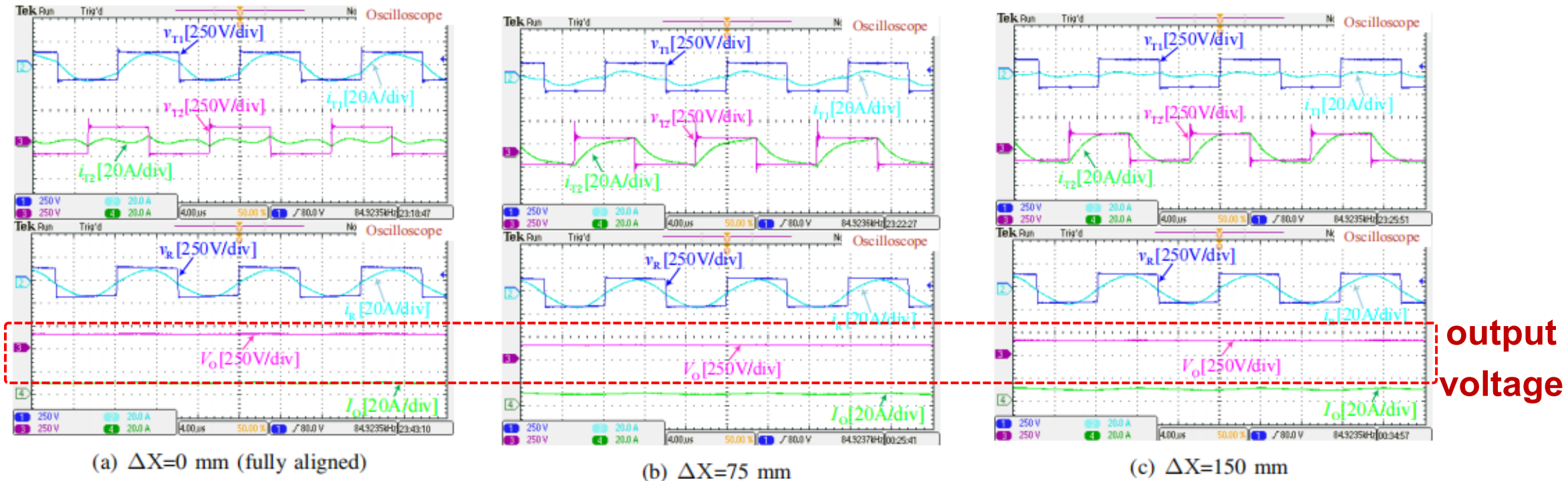
Experimental prototype of 1 kW proposed control-free IPT system

TABLE II
MEASURED PARAMETERS OF THE IPT PROTOTYPE

Coupler parameters		Circuit parameters	
Symbols	Value	Symbols	Value
L_{T1}	101.9 μH	L_1, C_1, C_{T1}	39.57 μH , 88.6 nF, 45.6 nF
L_{T2}	104.5 μH	L_2, C_2, C_{T2}	20.86 μH , 168.1 nF, 40.1 nF
L_R	102.2 μH	C_R	34.4 nF
M_{fa}	39.12 μH	R_L	22.5 Ω
N	10	V_{DC}	150 V
N_S	9	$D_1 - D_4$	MBR20200CTG
h_{gap}	60mm	$Q_1 - Q_8$	STC 4050

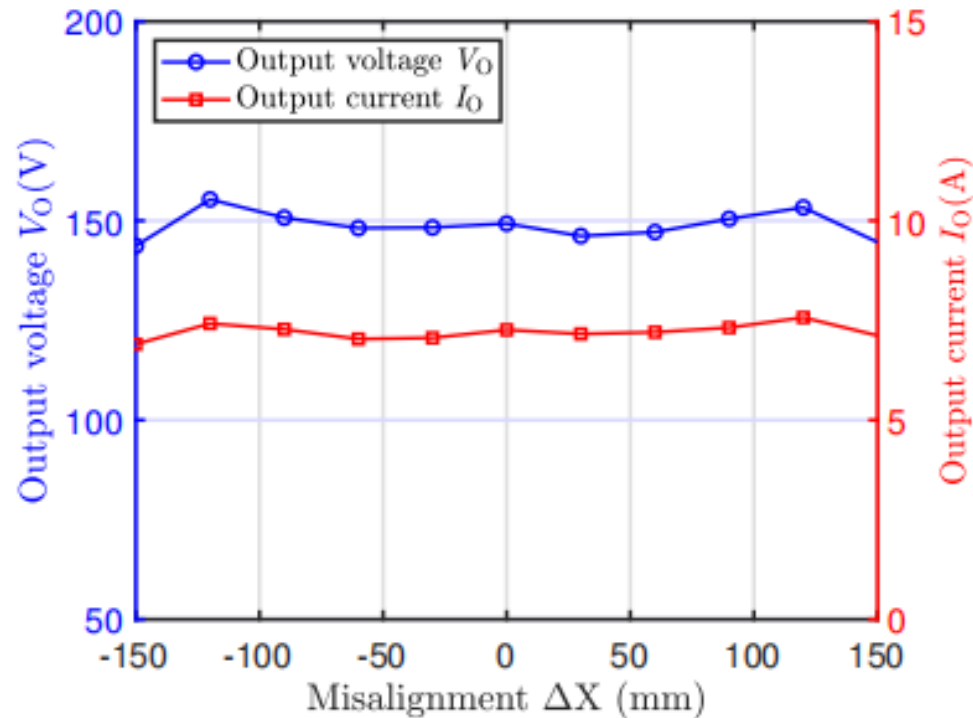
Measure parameters of IPT prototype

Experimental Validation

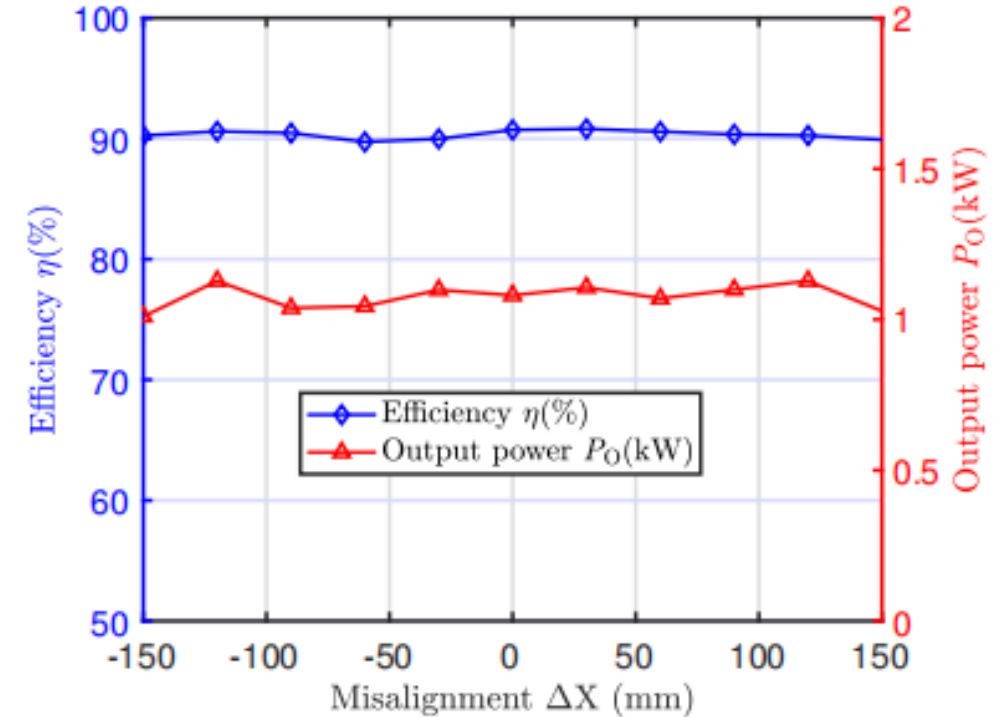


Experimental waveforms of the proposed control-free IPT system at various misalignment conditions

Experimental Validation



Measured output voltage and current versus misalignment along X-axis

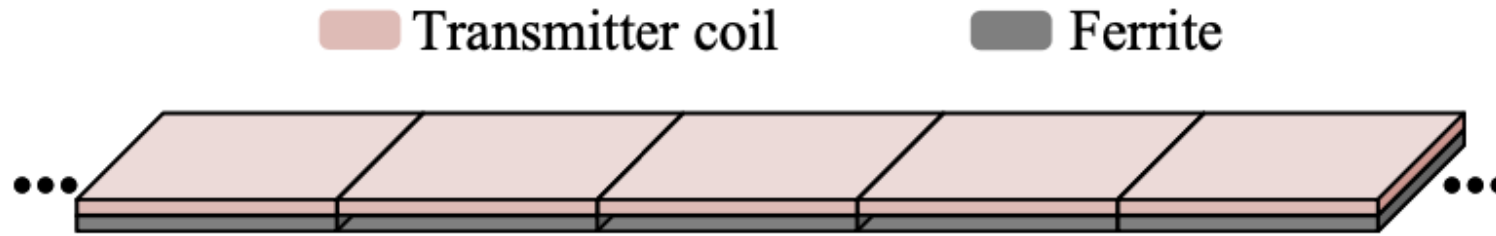


Measured DC-to-DC efficiency and output power versus misalignment along X-axis

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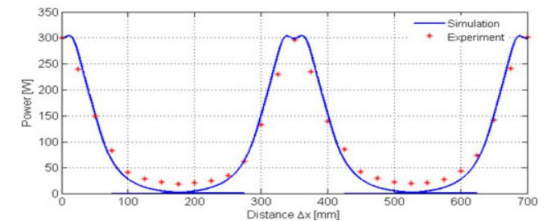
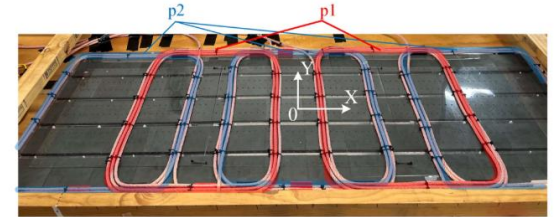
Dynamic Tx Track with Large Spacing Distance



(a) Without spacing distance

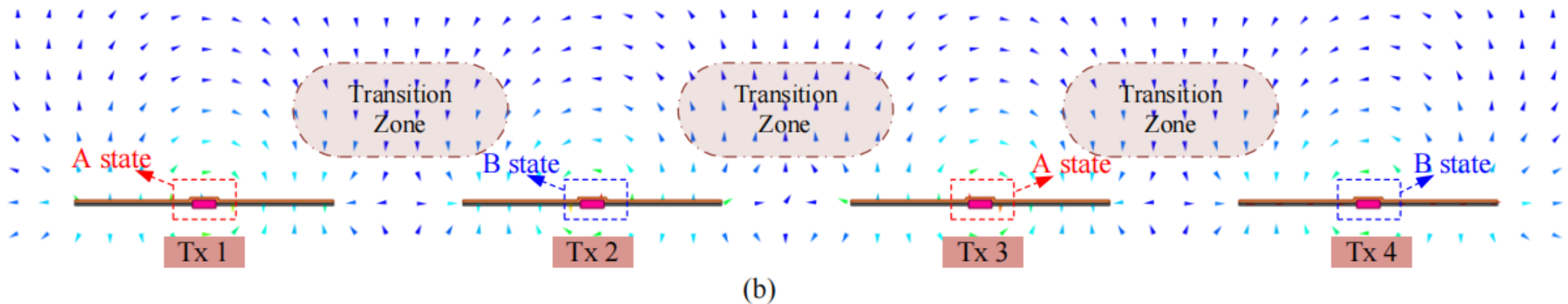
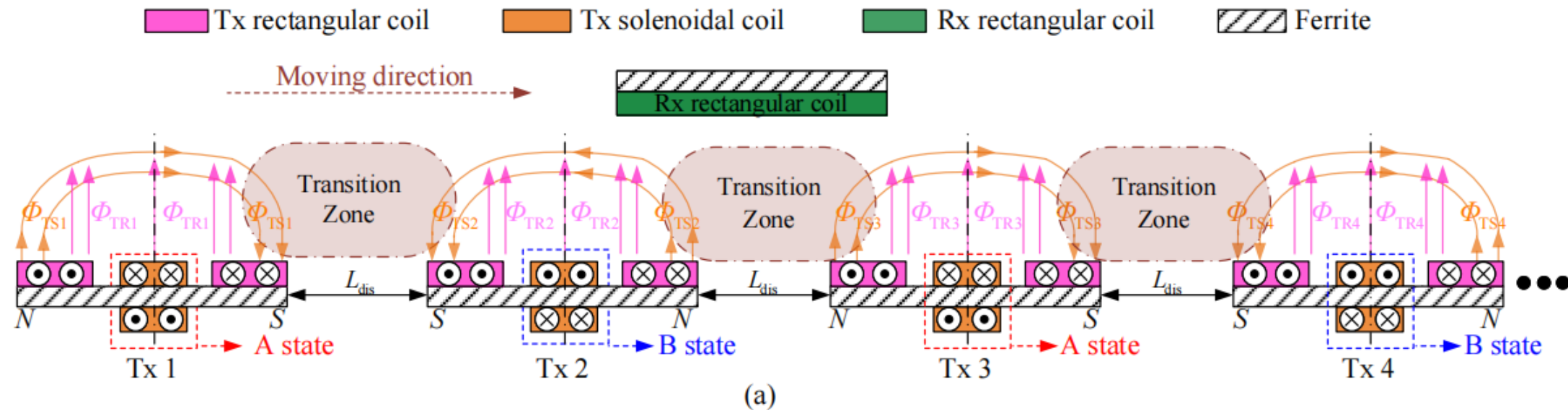


(b) With spacing distance



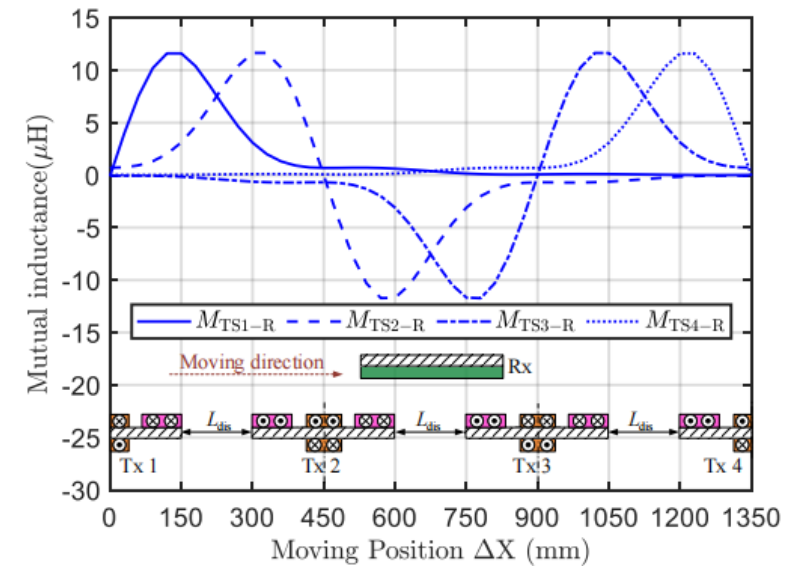
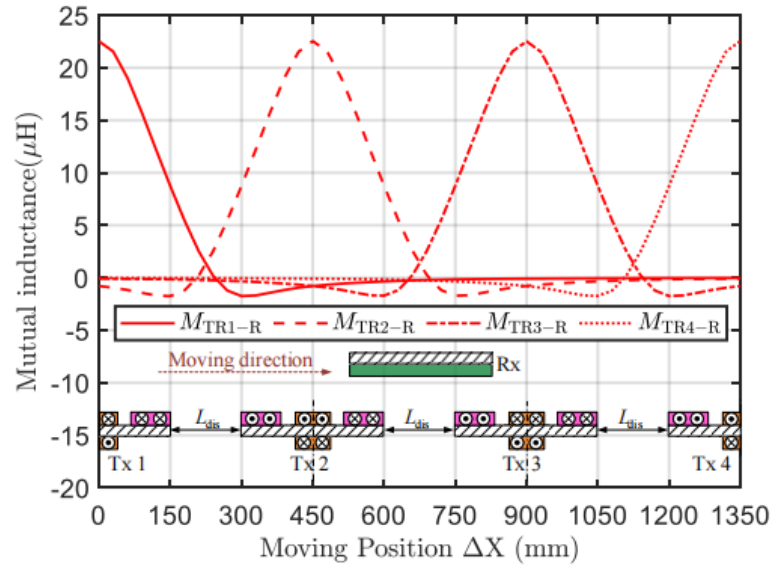
How to effectively mitigate output fluctuation while allowing large spacing distance?

Alternately-Arranged Segmented Transmitter Pads with Magnetic Field Complementation



Front view of the proposed DWPT Tx track consisting of
alternately-arranged RSPs with 50% spacing distance

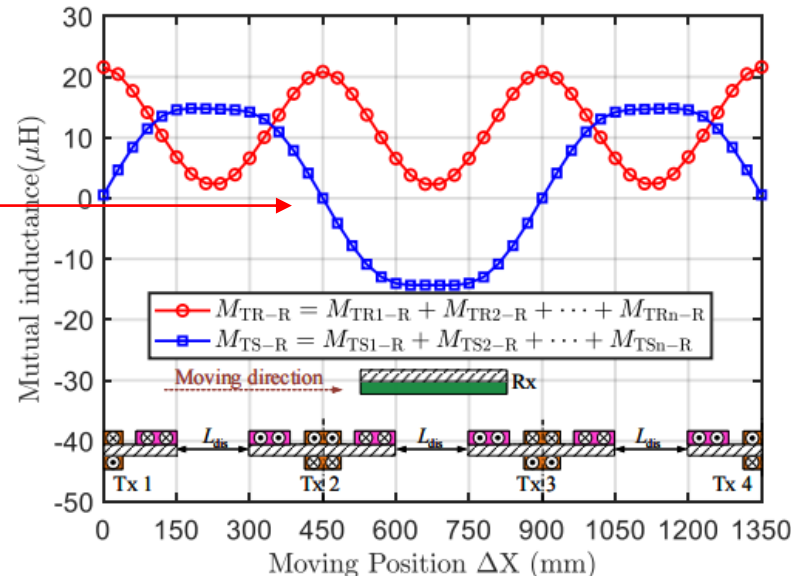
Operating Principle



Mutual inductances between Rx
RP and individual Tx RP coils

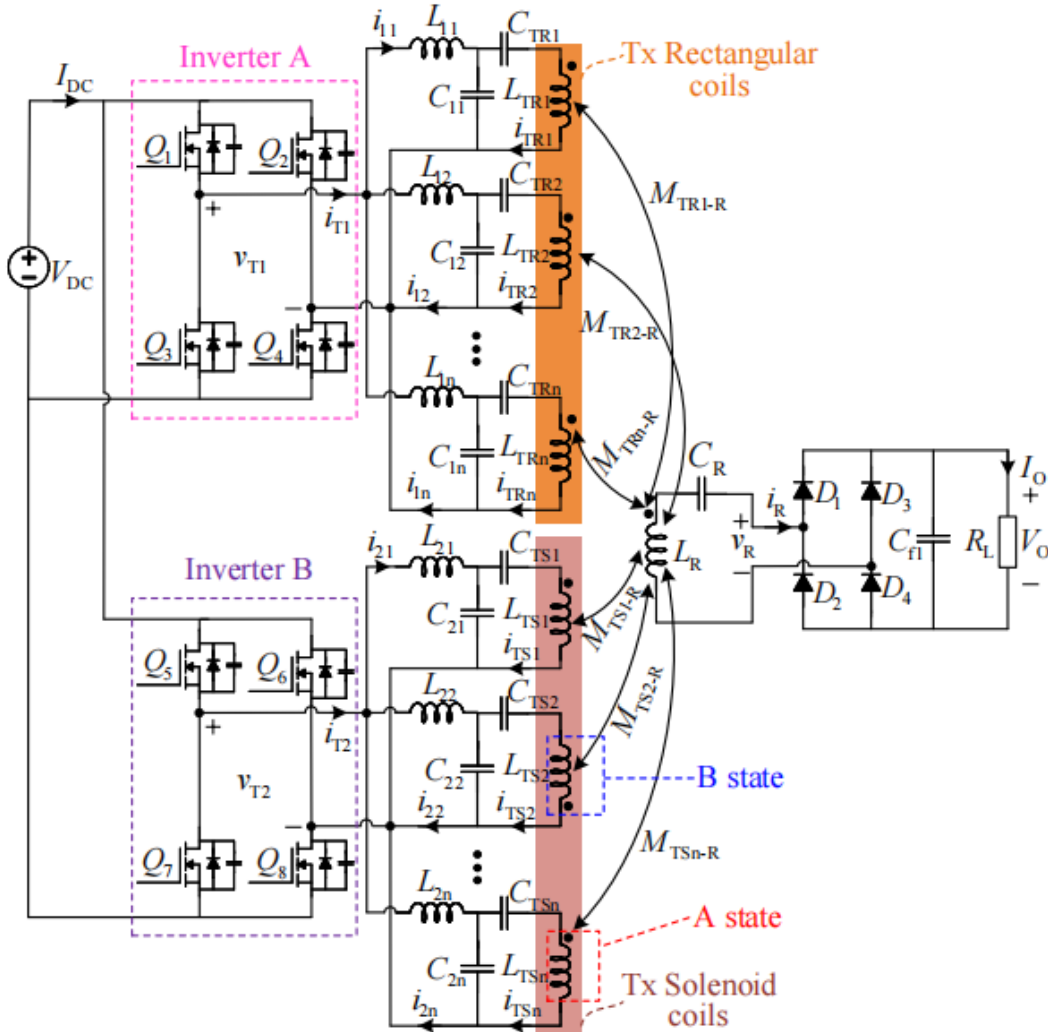
Mutual inductances between Rx
RP and individual Tx solenoid coils

Polarity Reversal Problem



Overall rectangular-to-rectangular and
solenoid-to-rectangular mutual
inductances

System Schematics and Circuit Modeling



$$\begin{aligned} V_{T1} &= \frac{4}{\pi} V_{DC} \angle 0^\circ \\ V_{T2} &= \frac{4}{\pi} V_{DC} \angle \theta^\circ, \end{aligned} \quad (4)$$

$$\begin{aligned} V_{T1} &= X_{11} I_{11} - X'_{11} I_{TR1} \\ 0 &= (X_{TR1} + X'_{11}) I_{TR1} - j\omega M_{TR1-R} I_R - X'_{11} I_{11} \\ V_{T1} &= X_{12} I_{12} - X'_{12} I_{TR2} \\ 0 &= (X_{TR2} + X'_{12}) I_{TR2} - j\omega M_{TR2-R} I_R - X'_{12} I_{12} \\ V_{T2} &= X_{21} I_{21} - X'_{21} I_{TS1} \\ 0 &= (X_{TS1} + X'_{21}) I_{TS1} - j\omega M_{TS1-R} I_R - X'_{21} I_{21} \\ V_{T2} &= X_{22} I_{22} - X'_{22} I_{TS2} \\ 0 &= (X_{TS2} + X'_{22}) I_{TS2} - j\omega M_{TS2-R} I_R - X'_{22} I_{22} \\ &\text{Equations for RSP3} \\ &\dots \\ &\text{Equations for RSPn} \\ 0 &= X_R I_R - \sum_{i=1}^n jX_{TRi-R} I_{TRi} - \sum_{i=1}^n jX_{TSi-R} I_{TSi} \end{aligned} \quad (5)$$

Circuit configuration of proposed DWPT system

Analysis of Output Function

$$V_O = \frac{\pi}{4} |\mathbf{V}_R| = \omega^2 V_{DC} C_1 \sqrt{\alpha + \beta}$$

Term ① $\approx$
constant



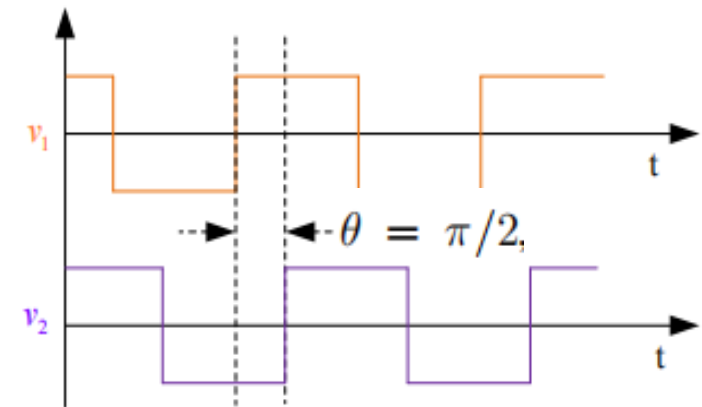
$$\alpha = \left(\sum_{i=1}^n M_{TRi-R} \right)^2 + \frac{C_2^2}{C_1^2} \left(\sum_{i=1}^n M_{TSi-R} \right)^2$$

$$\beta = \frac{2C_2 \sum_{i=1}^n M_{TRi-R} \sum_{i=1}^n M_{TSi-R} \cos\theta}{C_1}$$



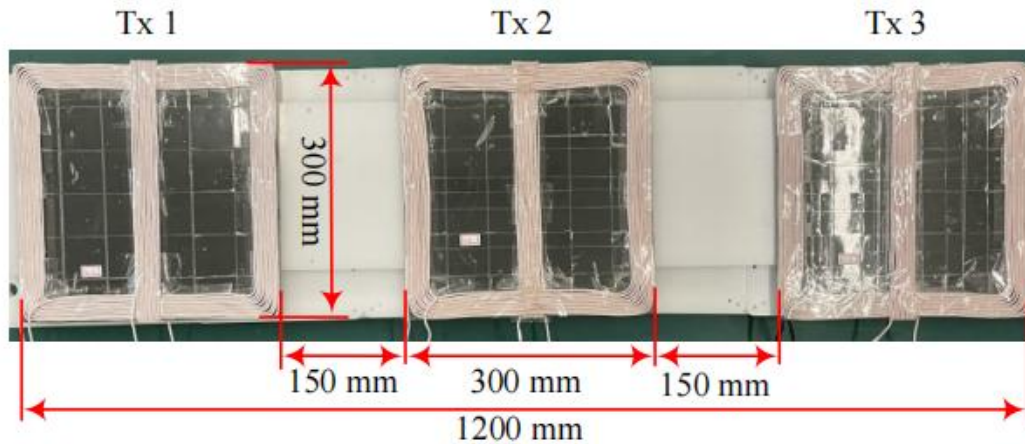
$$\Delta = M_{T1R}^2 + \frac{C_2^2}{C_1^2} M_{T2R}^2$$

← Term ② = 0



Orthogonal excitation

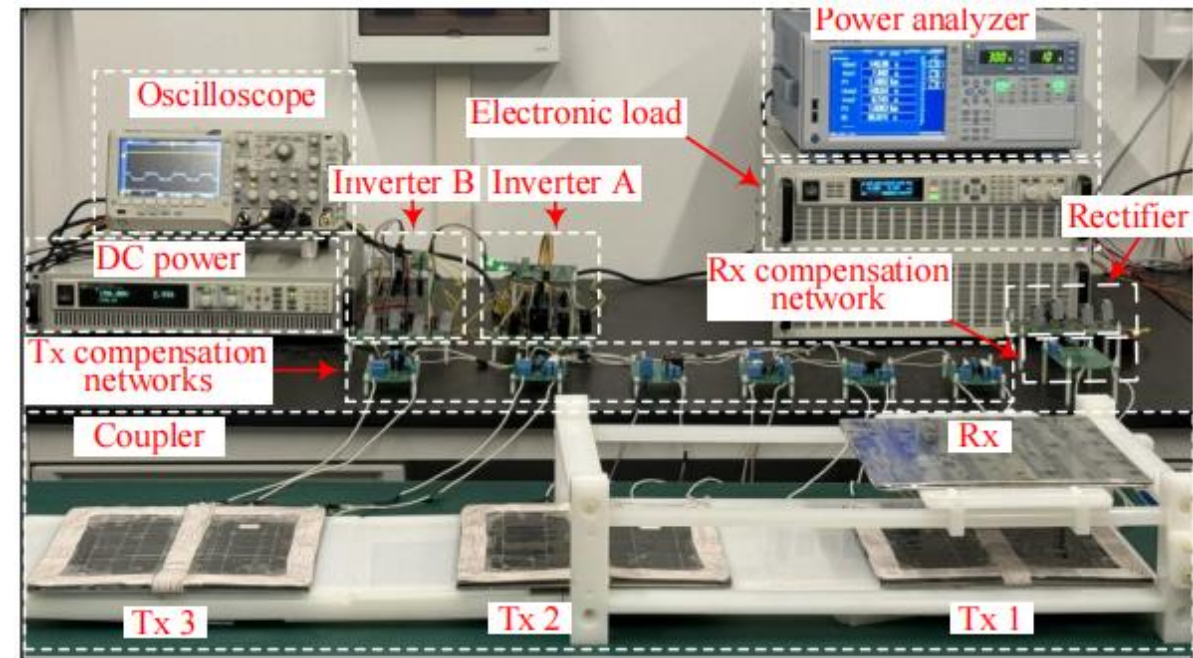
Experimental Validation



Practical structure and dimension of proposed DWPT transmitter

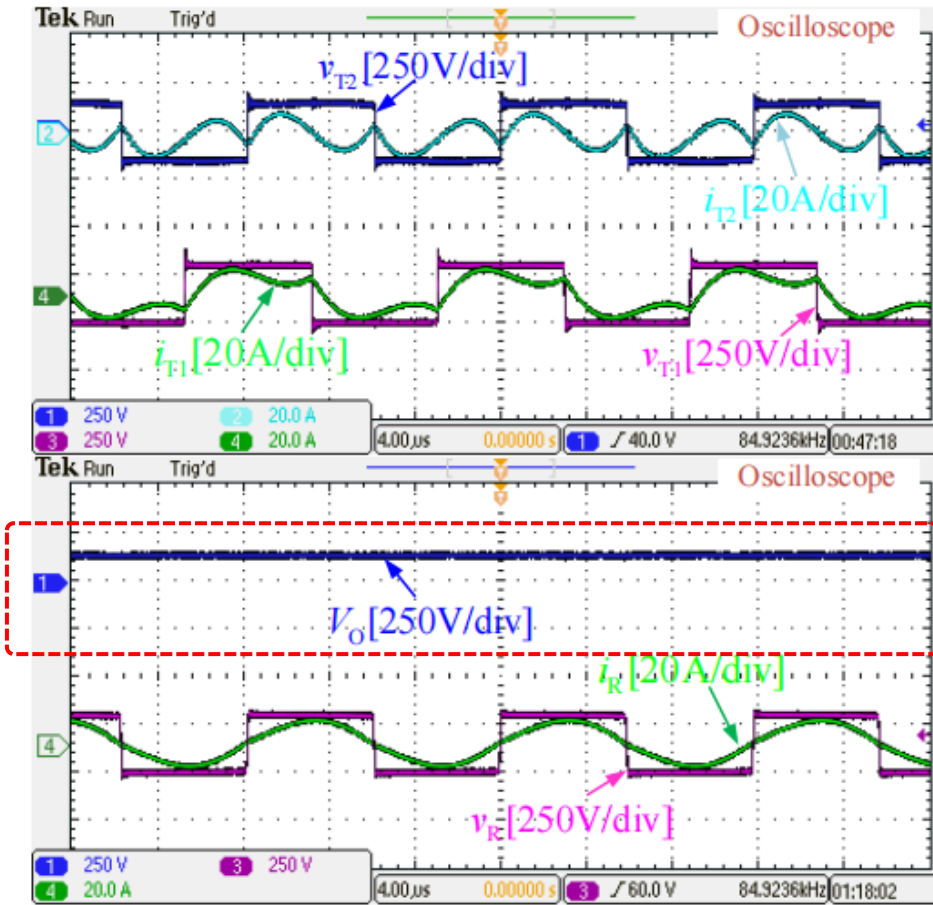
TABLE III
MEASURED PARAMETERS OF THE IPT PROTOTYPE

Coupler parameters		Circuit parameters	
Symbols	Value	Symbols	Value
L_{TRi} ($i=1,2,3$)	99.82 μH , 99.56 μH , 99.21 μH	L_1, L_2	21.65 μH , 15.32 μH
L_{TSi} ($i=1,2,3$)	101.81 μH , 100.98 μH , 101.36 μH	C_1, C_2	161.9 nF, 228.95 nF
L_R	99.86 μH	C_R	35.18 nF
h_{gap}	100 mm	C_{TR}, C_{TS}	41.96 nF, 42.06 nF
N	10	V_{DC}	150 V
N_S	9	$D_1 - D_4$	MBR20200CTG
	3 mm	$Q_1 - Q_8$	STC 4050

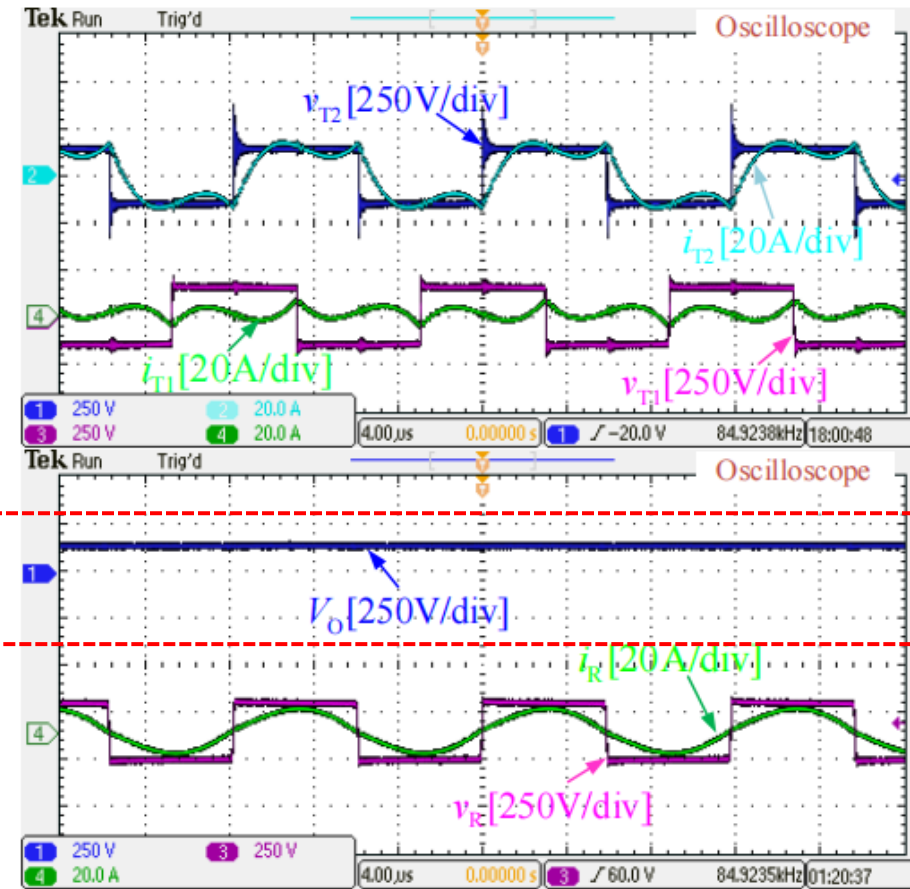


Experimental prototype of 1 kW proposed DWPT system with alternately-arranged segmented transmitters

Experimental Validation



(a) $\Delta X = 0$ mm

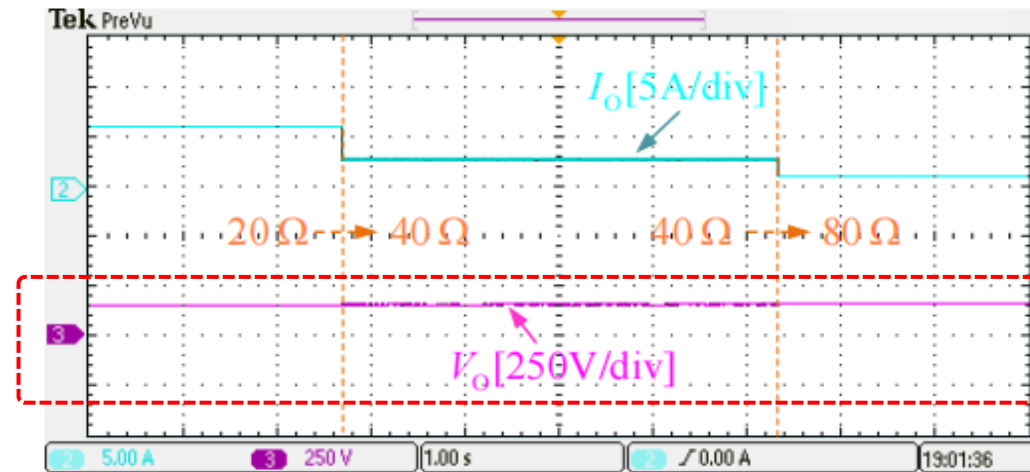


(b) $\Delta X = 675$ mm

output
voltage

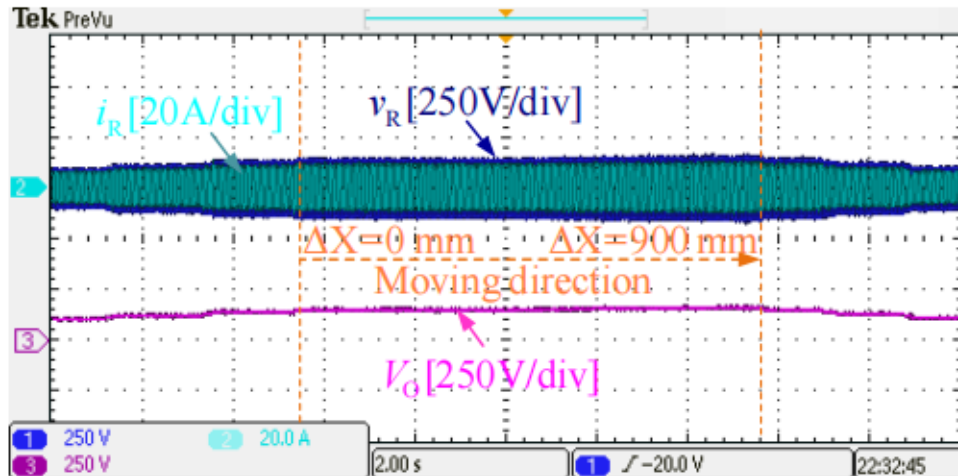
Experimental Validation

Output currents and voltages at different load conditions

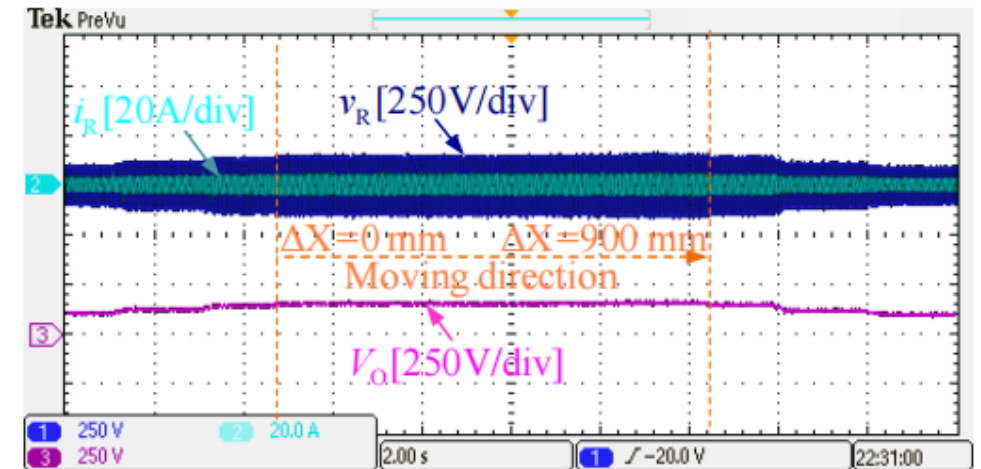


output voltage

Dynamic output currents and voltages at different load conditions

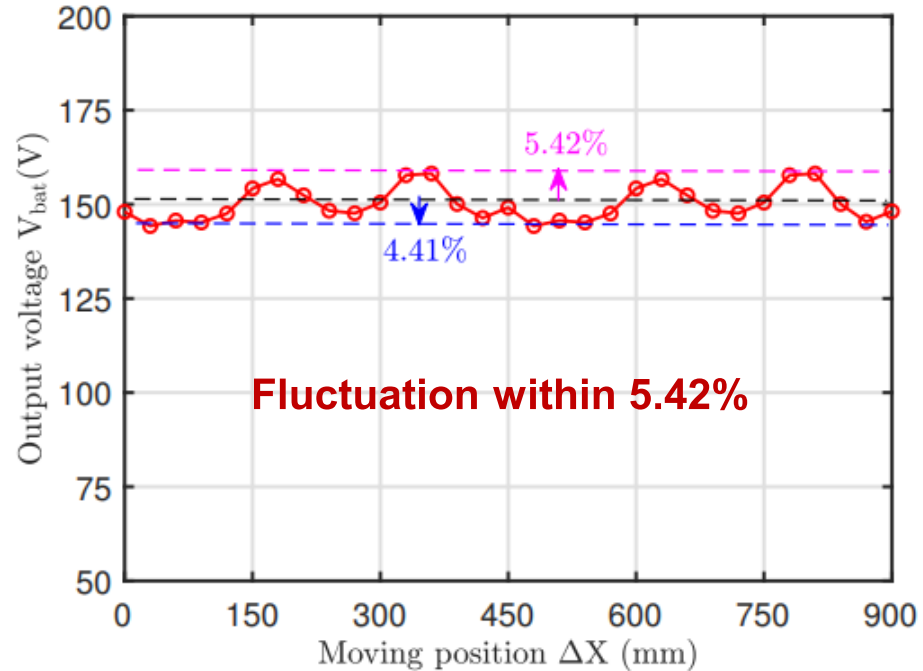


$R_L = 22.5\Omega$

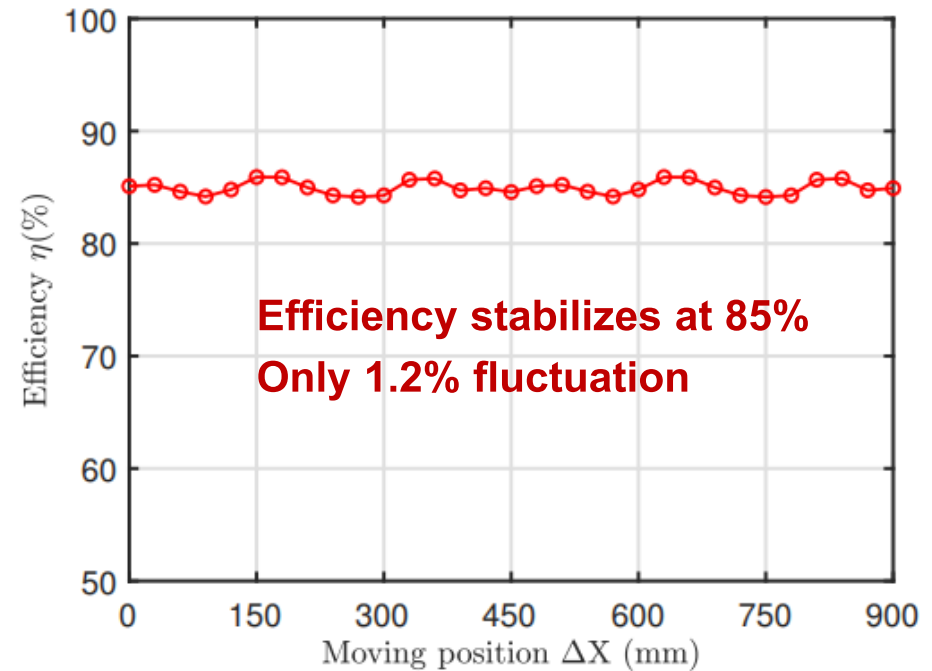


$R_L = 45\Omega$

Experimental Validation



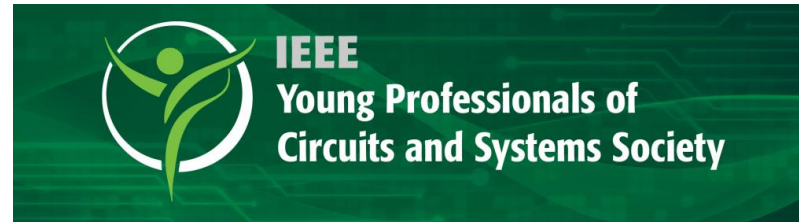
Output voltage V_O at full load during moving range $\Delta X \in [0 \text{ mm } 900 \text{ mm}]$



Measured dc-dc efficiency of proposed DWPT system within moving range $\Delta X \in [0 \text{ mm } 900 \text{ mm}]$ at full load

Conclusion

Our recent works proposed a novel rectangular solenoidal pad (RSP) which is a good approach to misalignment tolerance with reduced copper usage. We also developed a passive scheme to excite the proposed RSP without necessities of detection and control. We furthered extended the proposed RSP from static to dynamic wireless power transfer scenarios.



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